

→ Later when we examine stability we will be concerned with the non-dominant poles  
 - remember each pole also causes a  $90^\circ$  phase shift in output

→ Breadboard capacitances?

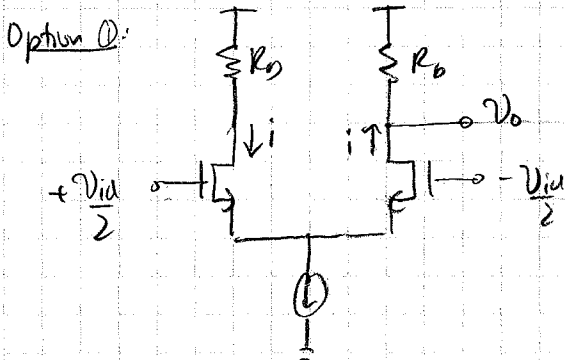
ECE 3110: Lecture #10

4- (Use in final project)  
 1- Long list  
 2- Quiz  
 3- homework comes

→ Back to differential pairs...

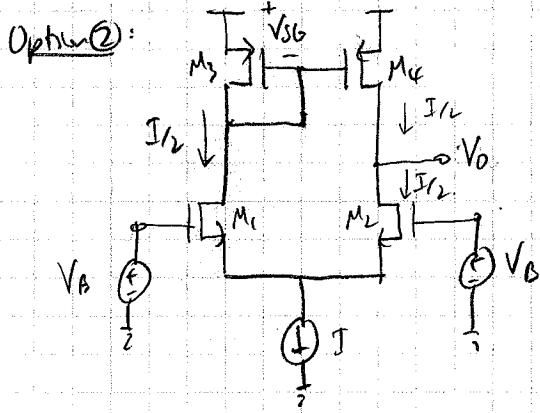
→ we saw that it is better to take a differential output signal from the pair  
 - better CMRR  
 - more gain ( $\times 2$ )

→ at some point the signal must be converted to single-ended



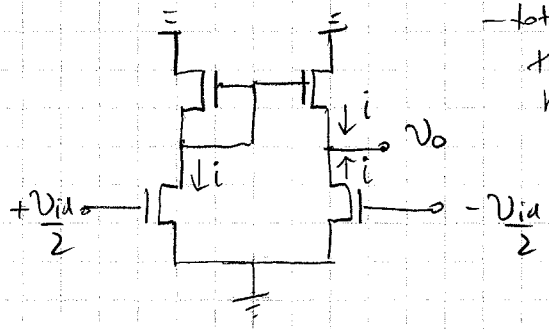
- just discard one output  
 - bad CMRR  
 - wasting gain

- We can do better, what about using an active load?



→ large signal:  $V_o = V_{DD} - V_{SG3}$

→ small signal:



- total current drawn to the output is  $2i$ , nothing is wasted.

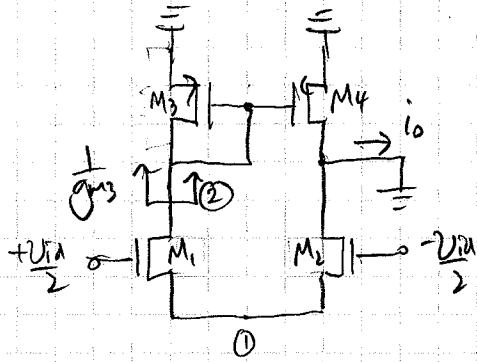
→ Analysis is a little more complicated than for resistive loads, since asymmetry prevents us from using the half circuit approach.

→ BUT... we will employ a "quite and dirty" shortcut from the book that simplifies things.

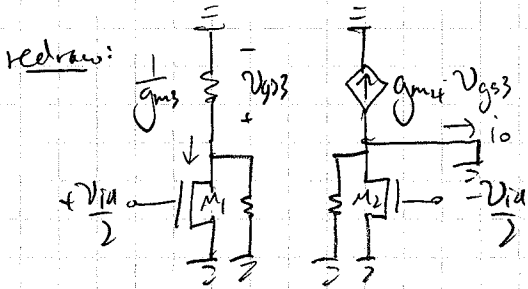
Differential Gain

- Find in two steps:
  - ① find the short-circuit transconductance
  - ② find the output resistance
- then multiply them together

① Short circuit transconductance: - ground output



- technically we can't make the tail circuit separated of a virtual ground at ①
- since at ② we see  $1/g_m$  to ground, if  $1/g_m$  is small, ② is "almost" grounded
- then assume symmetry and ① is a virtual ground.



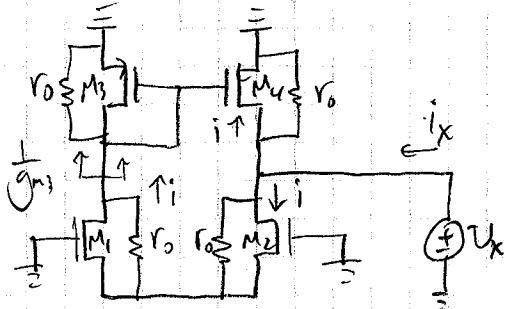
$\therefore$  in  $M_1$ : induces a current of  $g_{m1} \cdot \frac{V_{id}}{2}$   
 $\rightarrow V_{gs3} = -g_{m1} \cdot \frac{V_{id}}{2} \cdot \frac{1}{g_{m3}}$

$$\begin{aligned} \rightarrow i_o &= -\frac{V_{id}}{2} \cdot g_{m2} + g_{m4} \cdot V_{gs3} \\ &= \frac{V_{id}}{2} \cdot g_{m2} + g_{m4} \cdot \frac{V_{id}}{2} \cdot \frac{g_{m1}}{g_{m3}} \\ &= \frac{V_{id}}{2} \cdot g_{m1} + \frac{V_{id}}{2} \cdot g_{m1} \\ &= g_{m1} \cdot V_{id} \end{aligned}$$

If  $M_1 = M_2, M_3 = M_4$ :

② Now, to find output voltage we need to know the output resistance.

- ground inputs, put test source at output.



$\rightarrow$  current flowing into  $M_2$  sees  $\sim 2r_{o2}$  (neglect  $1/g_{m3}$ )

$$\therefore i = \frac{V_x}{2r_o}$$

$$\begin{aligned} \rightarrow \text{total } i_x &= 2 \cdot i + \frac{V_x}{r_{o4}} \\ &= 2 \cdot \frac{V_x}{2 \cdot r_{o2}} + \frac{V_x}{r_{o4}} \\ &= \frac{V_x}{r_{o2}} + \frac{V_x}{r_{o4}} \Rightarrow V_x = i_x \cdot R_o \end{aligned}$$

$$\rightarrow R_o = (r_{o2} \parallel r_{o4})$$

$\therefore$  Differential gain:  $A_d = g_{m1} \cdot (r_{o2} \parallel r_{o4})$