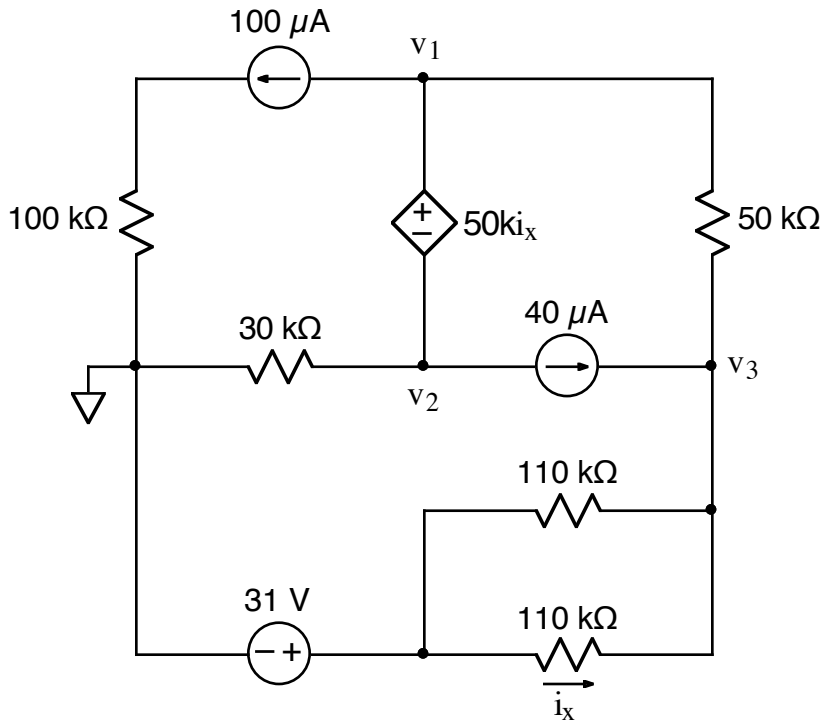


Ex:



Use the node-voltage method to find  $v_1$ ,  $v_2$ , and  $v_3$ .

**SOL'N:** Since we have a dependent source, we start by defining the dependent current in terms of the node voltages.

$$i_x = \frac{31 \text{ V} - v_3}{110 \text{ k}\Omega}$$

We have a supernode for nodes  $v_1$  and  $v_2$ . We first sum all the currents out of a bubble drawn around nodes  $v_1$  and  $v_2$  with the dependent voltage source inside:

$$100 \mu\text{A} + \frac{v_1 - v_3}{50 \text{ k}\Omega} + \frac{v_2}{30 \text{ k}\Omega} + 40 \mu\text{A} = 0 \text{ A} \quad (1)$$

We also write a voltage equation for the supernode:

$$v_1 - v_2 = 50 \text{ k}\Omega \cdot i_x = \frac{31 \text{ V} - v_3}{110 \text{ k}\Omega} \quad (2)$$

For the  $v_3$  node we have a conventional current sum:

$$\frac{v_3 - v_1}{50 \text{ k}\Omega} - 40 \text{ }\mu\text{A} + \frac{v_3 - 31 \text{ V}}{55 \text{ k}\Omega} = 0 \text{ A} \quad (3)$$

We clean up the first equation:

$$v_1 \frac{1}{50 \text{ k}\Omega} + v_2 \frac{1}{30 \text{ k}\Omega} - v_3 \frac{1}{50 \text{ k}\Omega} = -140 \text{ }\mu\text{A}$$

or, after multiplying by the common denominator, 150 k $\Omega$ :

$$3v_1 + 5v_2 - 3v_3 = -21 \text{ V} \quad (1')$$

And we clean up the third equation:

$$-v_1 \frac{1}{50 \text{ k}\Omega} + v_3 \frac{1}{50 \text{ k}\Omega} + \frac{1}{55 \text{ k}\Omega} = 40 \text{ }\mu\text{A} + \frac{31 \text{ V}}{55 \text{ k}\Omega}$$

or, after multiplying by the common denominator, 550 k $\Omega$ :

$$-11v_1 + 21v_3 = 22 + 310 \text{ V} = 332 \text{ V} \quad (3')$$

The simplest equation is Eqn (2), which we may solve for  $v_2$ :

$$v_2 = v_1 - 50 \text{ k}\Omega \frac{31 \text{ V} - v_3}{110 \text{ k}\Omega} = v_1 + \frac{5}{11} v_3 - \frac{155}{11} \text{ V}$$

Substituting this into Eqns (1') yields the following:

$$3v_1 + 5 \left( v_1 + \frac{5}{11} v_3 - \frac{155}{11} \text{ V} \right) - 3v_3 = -21 \text{ V}$$

or, after multiplying by 11:

$$88v_1 - 8v_3 = -21(11) \text{ V} + 775 \text{ V} = 544 \text{ V}$$

or, after dividing by 8:

$$11v_1 - v_3 = 68 \text{ V} \quad (1'')$$

Since  $v_2$  is absent from Eqn (3'), we proceed to solve (1'') and (3') for  $v_3$ .

(We solve for  $v_3$  because  $v_1$  is easier to eliminate.) We use (1'') + (3'):

$$\begin{array}{r} -11v_1 + 21v_3 = 332 \text{ V} \\ + 11v_1 - v_3 = 68 \text{ V} \\ \hline = 20v_3 = 400 \text{ V} \end{array}$$

or

$$v_3 = 20 \text{ V}$$

Using (1'') we find the value of  $v_1$  from  $v_3$ :

$$11v_1 = 68 \text{ V} + v_3 = 68 \text{ V} + 20 \text{ V} = 88 \text{ V}$$

or

$$v_1 = 8 \text{ V}$$

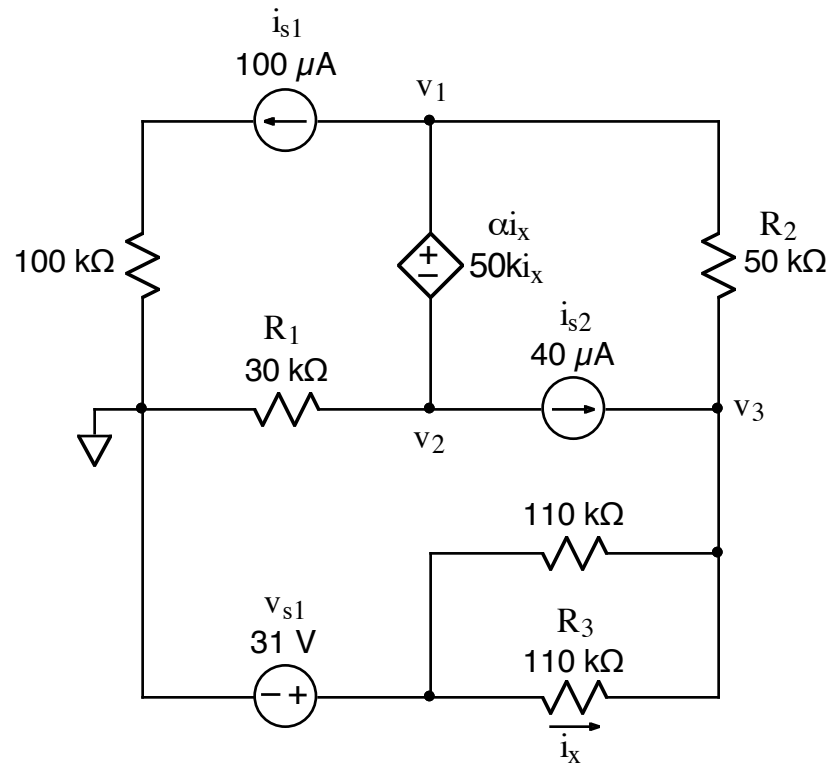
Finally, we use (1') to find  $v_2$ :

$$5v_2 = -21 \text{ V} - 3v_1 + 3v_3 = -21 \text{ V} - 3(8 \text{ V}) + 3(20 \text{ V}) = 15 \text{ V}$$

or

$$v_2 = \frac{15 \text{ V}}{5} = 3 \text{ V}$$

An alternative approach is to find a symbolic solution. We start by labeling the components in the circuit, as shown below:



Using conductance, (i.e., 1/resistance), we have the following symbolic equations:

$$i_{s1} + (v_1 - v_3)g_2 + v_2g_1 + i_{s2} = 0 \text{ A}$$

$$v_1 - v_2 = \alpha(v_{s1} - v_3)g_3$$

$$(v_3 - v_1)g_2 - i_{s2} + (v_3 - v_{s1})2g_3 = 0 \text{ A}$$

In matrix form, we have the following equation:

$$\begin{bmatrix} g_2 & g_1 & -g_2 \\ 1 & -1 & \alpha g_3 \\ -g_2 & 0 & g_2 + 2g_3 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} -(i_{s1} + i_{s2}) \\ \alpha v_{s1}g_3 \\ i_{s2} + 2v_{s1}g_3 \end{bmatrix}$$

If we multiply the second row by  $g_2$ , we have a convenient form for eliminating  $v_1$ :

$$\begin{bmatrix} g_2 & g_1 & -g_2 \\ g_2 & -g_2 & \alpha g_2 g_3 \\ -g_2 & 0 & g_2 + 2g_3 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} -(i_{s1} + i_{s2}) \\ \alpha g_2 g_3 v_{s1} \\ i_{s2} + 2v_{s1}g_3 \end{bmatrix}$$

We add the first and third rows, and we add the second and third rows to eliminate  $v_1$ . The result is a smaller matrix equation:

$$\begin{bmatrix} g_1 & 2g_3 \\ -g_2 & \alpha g_2 g_3 + g_2 + 2g_3 \end{bmatrix} \begin{bmatrix} v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} -i_{s1} + 2v_{s1}g_3 \\ \alpha g_2 g_3 v_{s1} + i_{s2} + 2v_{s1}g_3 \end{bmatrix}$$

Now can multiply the first row by  $g_2$  and the second row by  $g_1$  to obtain terms in the first column that will cancel out when the rows are summed:

$$\begin{bmatrix} g_1 g_2 & 2g_2 g_3 \\ -g_1 g_2 & [\alpha g_2 g_3 + g_2 + 2g_3]g_1 \end{bmatrix} \begin{bmatrix} v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} [-i_{s1} + 2v_{s1}g_3]g_2 \\ [\alpha g_2 g_3 v_{s1} + i_{s2} + 2v_{s1}g_3]g_1 \end{bmatrix}$$

Summing the first two rows yields an equation for  $v_3$ :

$$\begin{aligned} \{[\alpha g_2 g_3 + g_2 + 2g_3]g_1 + 2g_2 g_3\}v_3 &= [\alpha g_2 g_3 v_{s1} + i_{s2} + 2v_{s1}g_3]g_1 \\ &\quad + [-i_{s1} + 2v_{s1}g_3]g_2 \end{aligned}$$

or

$$v_3 = \frac{[\alpha g_2 g_3 v_{s1} + i_{s2} + 2v_{s1} g_3] g_1 + [-i_{s1} + 2v_{s1} g_3] g_2}{\{[\alpha g_2 g_3 + g_2 + 2g_3] g_1 + 2g_2 g_3\}}$$

This result looks daunting, but we can rearrange things somewhat:

$$v_3 = \frac{v_{s1}[(\alpha g_2 + 2)g_1 g_3 + 2g_2 g_3] - i_{s1} g_2 + i_{s2} g_1}{(\alpha g_2 + 2)g_1 g_3 + g_1 g_2 + 2g_2 g_3}$$

Multiplying the top and bottom by  $R_1 R_2 R_3$  simplifies the form somewhat:

$$v_3 = \frac{v_{s1}[(\alpha g_2 + 2)R_2 + 2R_1] - i_{s1} R_1 R_3 + i_{s2} R_2 R_3}{(\alpha g_2 + 2)R_2 + R_3 + 2R_1}$$

Now we calculate the various terms:

$$\alpha g_2 + 2 = 50 \text{ k}\Omega \cdot \frac{1}{50 \text{ k}\Omega} + 2 = 1 + 2 = 3$$

$$i_{s1} R_1 R_3 = 100 \text{ }\mu\text{A} \cdot 30 \text{ k}\Omega \cdot 110 \text{ k}\Omega = 3 \text{ V} \cdot 110 \text{ k}\Omega$$

$$i_{s2} R_2 R_3 = 40 \text{ }\mu\text{A} \cdot 50 \text{ k}\Omega \cdot 110 \text{ k}\Omega = 2 \text{ V} \cdot 110 \text{ k}\Omega$$

Substituting for all terms, we obtain a value for  $v_3$ :

$$v_3 = \frac{31 \text{ V}[3 \cdot 50 \text{ k}\Omega + 2 \cdot 30 \text{ k}\Omega] - 3 \text{ V} \cdot 110 \text{ k}\Omega + 2 \text{ V} \cdot 110 \text{ k}\Omega}{3 \cdot 50 \text{ k}\Omega + 110 \text{ k}\Omega + 2 \cdot 30 \text{ k}\Omega}$$

or

$$v_3 = \frac{31 \text{ V}[210 \text{ k}\Omega] - 1 \text{ V} \cdot 110 \text{ k}\Omega}{320 \text{ k}\Omega} = \frac{31 \text{ V}[21] - 1 \text{ V} \cdot 11}{32}$$

or

$$v_3 = \frac{640 \text{ V}}{32} = 20 \text{ V}$$

From the top row of the first 2 x 2 matrix equation, above, we obtain an equation relating  $v_2$  and  $v_3$ :

$$g_1 v_2 + 2g_3 v_3 = -i_{s1} + 2v_{s1} g_3$$

$$\begin{bmatrix} g_1 & 2g_3 \\ -g_2 & \alpha g_2 g_3 + g_2 + 2g_3 \end{bmatrix} \begin{bmatrix} v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} -i_{s1} + 2v_{s1} g_3 \\ \alpha g_2 g_3 v_{s1} + i_{s2} + 2v_{s1} g_3 \end{bmatrix}$$

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or

$$v_2 = \frac{-i_{s1} + 2v_{s1}g_3 - 2g_3v_3}{g_1}$$

Multiplying top and bottom by  $R_1R_3$ , we have the following result:

$$v_2 = \frac{-i_{s1}R_1R_3 + 2v_{s1}R_1 - 2R_1v_3}{R_3}$$

or

$$v_2 = \frac{-100 \mu\text{A} \cdot 30 \text{ k}\Omega \cdot 110 \text{ k}\Omega + 2 \cdot 31 \text{ V} \cdot 30 \text{ k}\Omega - 2 \cdot 30 \text{ k}\Omega \cdot 20 \text{ V}}{110 \text{ k}\Omega}$$

or

$$v_2 = \frac{-3 \text{ V} \cdot 110 \text{ k}\Omega + 62 \text{ V} \cdot 30 \text{ k}\Omega - 40 \text{ V} \cdot 30 \text{ k}\Omega}{110 \text{ k}\Omega}$$

or

$$v_2 = \frac{-3 \text{ V} \cdot 11 + 62 \text{ V} \cdot 3 - 40 \text{ V} \cdot 3}{11} = \frac{3 \text{ V} \cdot 11}{11} = 3 \text{ V}$$

Finally, we return to our original equation for the supernode voltage relationship to find  $v_1$ :

$$v_1 - v_2 = \alpha(v_{s1} - v_3)g_3$$

or

$$v_1 = 50 \text{ k}\Omega(31 \text{ V} - 20 \text{ V})/110 \text{ k}\Omega + 3 \text{ V} = 8 \text{ V}$$

To verify the above answers, we perform a consistency check. Using the node voltages, we calculate the currents flowing in the circuit and check to see if they sum to zero at each node.

We find  $i_x$  using  $v_3$  and the 31 V source.

$$i_x = \frac{31 \text{ V} - 20 \text{ V}}{110 \text{ k}\Omega} = \frac{11 \text{ V}}{110 \text{ k}\Omega} = 100 \mu\text{A}$$

We find the other currents by using Ohm's law. The diagram below shows that the currents at every node sum to zero. It follows that our solution is correct.

