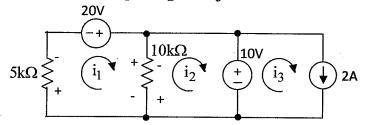


Use the mesh-current method to find i_1 and i_2 , and i_3 .



Procedure for solving for mesh-currents:

- 1. Label every mesh with a mesh current variable.
- 2. Label polarity on every R. If R is between 2 meshes, there will be polarities labeled for two mesh currents.
- 3. If an outside branch contains a current source (dependent or independent), the mesh current is equal to that current source value (+ if they are in the same direction or - if opposite directions).
- 4. If the circuit contains a dependent source, the dependent variable needs to be written in terms of the mesh currents.
- 5. If a current source is located between two meshes, a supermesh equation is written. The supermesh equation is the current source value equal to the two mesh currents going through it. (+ if the same direction, - if opposite directions).
- 6. A voltage loop is taken through pathways without current sources in them. The voltage loop equation will contain two currents going through a resistor located between 2 meshes.
- 7. Simultaneous equations are solved to obtain the mesh current values. These mesh current values can be used to determine any other unknown variable in the circuit.

Step 1 and 2 are shown in the figure above:

Step 3:
$$i_3 = 2A$$

$$-i_1(5k) + 20 - i_1(10k) + i_2(10k) = 0$$
$$+i_1(10k) - i_2(10k) - 10 = 0$$
$$-i_1(5k) + 20 - 10 = 0$$

The last equation only has i₁ as an unknown, solving this gives:

$$-i_1(5k) + 20 - 10 = 0$$

 $i_1(5k) = 10$ $\Rightarrow i_1 - 2m$

$$i_1(5k) = 10 \implies i_1 = 2mA$$

Using the second equation and this known value yields:

$$+i_2(10k) = +i_1(10k)-10$$

$$+i_2 = +i_1 - 1m$$

$$+i_2 = 2m - 1m = 1mA$$

$$i_1 = 2mA$$
 $i_2 = 1mA$ $i_3 = 2A$



- 2. a. Use the mesh-current method to find V_x , V_x must not be in equation.
 - b. Find power dissipated by the dependent source.

Step 1 and 2 are shown in the figure at the right:

Step 3:
$$i_1 = -50mA$$
, $i_3 = +0.5A$

Step 4: Vx is the voltage across the current source and the 500 ohm:

$$V_x = (+i_3 - i_2)500 = (+0.5 - i_2)500 = +250 - i_2500$$

Step 5: No supermesh

Step 6(only one voltage loop possible):

$$+2V_x - i_2 1k + i_1 1k + 7 + V_x = 0$$

$$3(+250-i_2500)-i_21k+50m(1k)+7=0$$

$$i_2 \cdot 2.5k = 807$$

$$i_2 = 282.8 mA$$

Plugging this value into the equation for Vx above gives:

$$V_x = +250 - (282.8m)500 = 108.6V$$

Power in the dependent source:

$$P = I * V = (\mathbf{i}_1 - \mathbf{i}_2)2(V_x) = (282.8m + (-50m))2(108.6) = -72.3W$$



Find V_{th} . This is the open circuit voltage between points a-b: (Using mesh currents)

$$i_1 = -5A$$

$$+20-i_15-i_15-i_13-i_12+i_22=0$$

$$20 - i_1 15 + i_2 2 = 0$$

$$20 - i_1 15 + (-5)2 = 0$$

$$i_1 = \frac{2}{3}$$

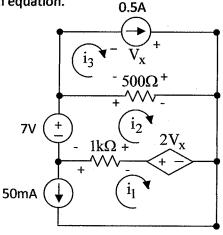
An equation for Vth is needed:

$$V_{th} - 0 - i_1(3) = 0$$

$$V_{th} = \frac{2}{3}(3) = 2V$$

Rth is found by removing all the independent sources (20V becomes a wire, 5A becomes an open):

$$R_{th} = 0.6 + 3 \parallel (10 + 2) = 0.6 + \frac{1}{\frac{1}{3} + \frac{1}{12}} = 3\Omega$$



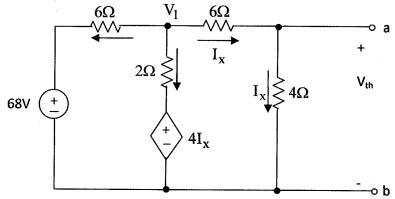
 0.6Ω

5A

 3Ω



4. Find the Thevenin equivalent circuit at terminals a-b.



Using Node-Voltage Method - Procedure:

- 1. a. Place a reference point (0V).
 - b. Label every essential node with a node-voltage.
- 2. Place all currents through every R.
- 3. If the circuit contains a dependent source, the dependent variable needs to be written in terms of the node voltages.
- 5. If only a single voltage source is located in a branch, a supernode equation is written. The supernode equation is the voltage source value equal to the two nodes on both sides of it. The voltage difference is taken as + at the positive side and at the negative sign.
- 6. A current summation is taken at all node voltage variables. Each current term is a positive on the + side of the labeled current and on the negative side of the labeled current.
- 7. Simultaneous equations are solved to obtain the node voltage values. These node voltage values can be used to determine any other unknown variable in the circuit.

In this case, the reference point is placed at b. The dependent variable is written in terms of V_1 .

$$I_x = \frac{V_1}{(6+4)} = \frac{V_1}{10}$$

There are no supernodes for this circuit, so a current summation is written:

$$\frac{\left(V_1 - 68\right)}{6} + \frac{\left(V_1 - 4\left(\frac{V_1}{10}\right)\right)}{2} + \frac{\left(V_1\right)}{10} = 0$$

$$V_1\left(\frac{1}{6} - \frac{1}{5} + \frac{1}{2} + \frac{1}{10}\right) = \frac{\left(68\right)}{6}$$

$$V_1 = \frac{\left(68\right)}{6} \cdot \left(\frac{30}{17}\right) = 20V$$

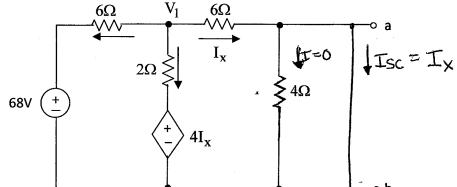
Using this solution for V₁:

$$I_x = \frac{V_1}{10} = \frac{20}{10} = 2A$$

Using ohm's law:



Find the Thevenin equivalent circuit at terminals a-b.



Solving using node-voltage: $T_x = \frac{V_1}{6}$ $\left(\frac{V_1-68}{6}\right) + \frac{V_1-4\left(\frac{V_1}{6}\right)}{2} + \frac{V_1}{6} = 0$

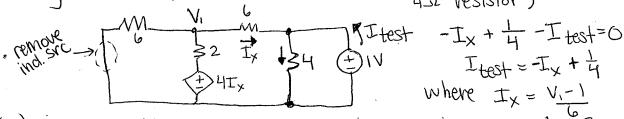
$$V_{1}\left(\frac{1}{6} + \frac{1}{2} - \frac{2}{6} + \frac{1}{6}\right) = \frac{68}{6}$$

$$V_{1} = \frac{68}{6}, \frac{6}{3} = \frac{68}{3}$$

$$I_{SC} = \frac{V_1}{6} = \frac{68}{3(6)} = \frac{68}{18}$$

R+h =
$$\frac{V_{+h}}{I_{sc}} = \frac{18}{68} \cdot (8) = \frac{144}{68} \cdot \frac{36}{17}$$

test source: (I used a vsrc to set a known current through Using



$$\frac{(V_1)}{6} + \frac{V_1 - 4(\frac{V_1}{6} - \frac{1}{6})}{2} + \frac{(V_1 - 1)}{6} = 0$$

$$V_1(\frac{1}{6} + \frac{1}{2} - \frac{2}{6} + \frac{1}{6}) = -\frac{1}{6}$$

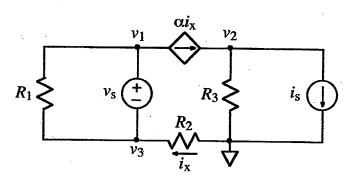
$$V_2 = -\frac{1}{6} \cdot \frac{6}{3} = -\frac{1}{3}$$



5. Determine the power in the dependent source if $R_L = 2k\Omega$ $V_1 \qquad 4k\Omega \qquad 2000I_X$ $V_2 \qquad V_3 \qquad V_4 \qquad V_4$

$$V_{1}(6k^{-1}6$$

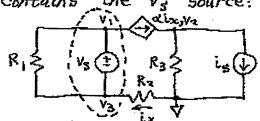
: power =
$$2000(\frac{8.2}{6k}) \cdot (.91m) = [+2.5mW]$$



- For the circuit shown, write three independent equations for the node voltages v_1 , v_2 , and v_3 . The quantity i_x must not appear in the equations.
- Make a consistency check on your equations for part 1(a) by setting resistors and sources to values for which the values of v_1 , v_2 , and v_3 are obvious. State the values of resistors, sources, and node voltages for your consistency check, and show that your equations for problem 1(a) are satisfied for these values. (In other words, plug the values into your equations for problem 1(a) and show that the left side and the right side of each equation are equal.)
 - 6. sol'n: Define ix in terms of node V's:

$$L_X = \frac{OV - V_3}{R_2} = -\frac{V_3}{R_2}$$

Now we write three egins for the node V's. We have a supernode for V, and V3, so we sum the currents out of a bubble around V, and V3 that contains the Vs source:



The sum of currents out of the bubble has two terms that cancel out:

$$\frac{v_1 - v_3}{R_1} + \alpha \left(\frac{-v_3}{R_2} \right) + \frac{v_3 - v_1}{R_1} + \frac{v_3}{R_2} = 0A$$

 $V_3\left(\frac{1}{R_2} - \frac{\alpha}{R_2}\right) = 0A \tag{1}$

We also get a voltage eg'n for the supernode:

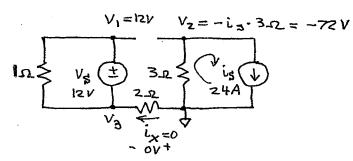
$$V_1 - V_3 = V_5 \tag{2}$$

Our third eg'n comes from node V2:

$$-\alpha\left(-\frac{v_3}{R_2}\right) + \frac{v_2}{R_3} + i_5 = 0A \qquad (3)$$

7. Many different consistency checks are possible. One example is given here.

Let $\alpha=0$, $V_3=12V$, $i_5=24A$ $R_1=1$, $R_2=2$, $R_3=3$.



We must have $i_x=0$ or we will accumulate charge on the left side of the circuit. Thus, $v_3=0V$.

Since the v_s source connects v_3 to v_i , we have $v_1 = v_3 + v_5 = 0V + 12V = 12V$.

On the right side, is flows around the loop, giving $v_z = -i_3 \cdot 3 \Omega = -24A \cdot 3\Omega = -72V$.

$$V_1 = 12V$$
, $V_2 = -72V$, $V_3 = 0V$

Now we plug these values into our 3 eg'ns from part (a):

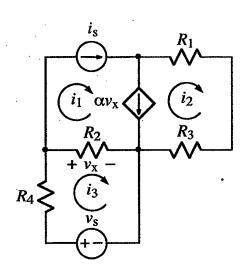
$$OV\left(\frac{1}{2\Omega} - \frac{O}{2\Omega}\right) \stackrel{?}{=} OA \qquad (1)$$
or
$$OA \stackrel{?}{=} OA \qquad (2)$$

$$|2v-o\gamma|^{\frac{2}{5}}|2V| \qquad (2)$$

$$-0\left(-\frac{12V}{2\Omega}\right) + \frac{-72V}{3\Omega} + 24A \stackrel{?}{=} 0A$$
or $-24A + 24A \stackrel{?}{=} 0A V$ (3)

The egins are all satisfied, completing the consistency check.

8



For the circuit shown, write three independent equations for the three mesh currents i_1 , i_2 , and i_3 . The quantity v_x must not appear in the equations.

soln: c) The loops for i, and iz share a current source, xvx, meaning they form a super-mesh. For the voltage loop around the outside of the i, and iz loops, we have the following problem: the i, loop includes current source is. Thus, we must abandon the outer voltage loop.

Instead, we have $i, = i \le 1$, since sre is is on the outside edge of the circuit. We need two egins for the two loops, and the other egin is the usual current source egin for the xv_x source in terms of i, and i. $xv_x = i_1 - i_2$

We must write vx in terms of mesh durrents, however

$$V_{x} = (i_3 - i_1) R_2$$

Using this expression for v_x , we obtain a second equation for the super-mesh:

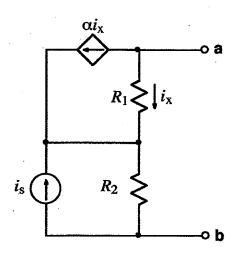
$$\left(i_3-i_1\right)R_2=i_1-i_2$$

The voltage loop for i3 gives the third and final equation:

$$-i_3R_4 - i_3R_2 + i_1R_2 + V_5 = 0V$$

Note: i, may be replaced by is in the preceding two eghs.

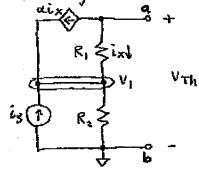
9



Find the Thevenin equivalent circuit at terminals a and b. i_x must not appear in your solution. Note: $\alpha > 0$.

soln:

VTh = Valb with nothing connected across terminals a and b. We can use the node-voltage method to find Vm.



we have one node: v₁. After defining ix in terms of v₁, we sum the currents out of the v₁ node. For ix, however, the only egin we can write is the following:

$$i_X = -\alpha i_X$$
 where $\alpha > 0$ or $i_X (1+\alpha) = 0A$ or $i_X = 0A$

So no current flows in either R₁ or the dependent source. Thus, our node-voltage eg'n for V, is as follows:

$$-is + \frac{V_1}{R_2} = 0A$$

or

$$v_1 = igR_2$$

Since $i_x = 0$, the voltage drop across R, is zero. It follows that

$$V_{\text{Th}} = V_1 = i_5 R_2$$

One way to find R_{Th} is to short a to b, find the current, isc, flowing in the short, and compute

$$R_{Th} = \frac{V_{Th}}{i_{SC}}.$$

$$\alpha i_{x} \text{ ov } \alpha$$

$$R_{1} > \downarrow i_{x}$$

$$V'_{1} \qquad \downarrow i_{SC}$$

$$i_{S} \Leftrightarrow R_{2} > 0$$

We use the node-voltage method again.

ix defined in terms of vi is

$$i_X = -\frac{v_1}{R_1}$$

The summation of currents out of the v, node is as follows:

$$-i_{S} + \frac{v_{1}'}{R_{2}} - \alpha \left(-\frac{v_{1}'}{R_{1}}\right) - -\frac{v_{1}'}{R_{1}} = OA$$

 $v_1'\left(\frac{1+\alpha}{R_1}+\frac{1}{R_2}\right)=is$

or
$$V_{1} = \frac{ig}{\frac{1+\kappa+1}{R_{1}R_{2}}} \cdot \frac{R_{1}R_{2}}{R_{1}R_{2}}$$
or
$$V_{1} = \frac{ig}{R_{1}R_{2}} \cdot \frac{R_{1}R_{2}}{R_{1}R_{2}}$$

$$V_{1} = \frac{ig}{(1+\kappa)R_{2}+R_{1}}$$

$$V_1' = \frac{R_1 R_2}{(1+\alpha)R_2 + R_1}$$

 $i_X = -\frac{v_1'}{R} = -\frac{i_S R_2}{(1+\alpha) R_2 + R_1}$

We find isc by summing currents out of the top node.

$$\alpha i_x + i_x + i_{sd} = OA$$

or

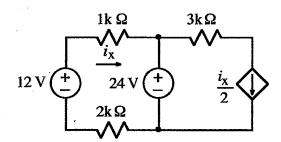
Thus,
$$R_{Th} = \frac{V_{Th}}{i_{sc}} = \frac{(1+\alpha)i_{s}R_{z}}{(1+\alpha)R_{z}+R_{1}}$$

$$\frac{V_{Th}}{i_{sc}} = \frac{(1+\alpha)R_{z}+R_{1}}{(1+\alpha)i_{s}R_{z}} = \frac{(1+\alpha)R_{z}+R_{1}}{1+\alpha}$$

$$R_{Th} = \frac{V_{Th}}{i_{sc}} = \frac{i_{s}R_{z}}{(1+\alpha)i_{s}R_{z}} = \frac{(1+\alpha)R_{z}+R_{1}}{1+\alpha}$$

$$\frac{(1+\alpha)R_{z}+R_{1}}{(1+\alpha)R_{z}+R_{1}}$$

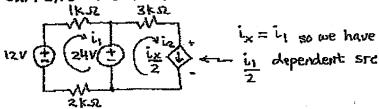
10.



Calculate the power consumed (i.e., dissipated) by the $i_x/2$ dependent source. Note: If a source supplies power, the power it consumes is negative.

soln: Any method of soln is acceptable.

Here, the soln employs the meshcurrent method.



in loop: 12V - in 1ks - 24V - in 2ks = OV

Solving for i, we have

$$i_1 = \frac{24V - 12V}{1kR + 2kR} = -\frac{12V}{3kR} = -\frac{4mA}{3}$$

We have $i_x = i_1$, so the loop current i_2 is $i_x/2 = i_1/2$. (Note that i_2 is the same as the current in the dependent src, which is on the outside edge of the circuit.)

$$i_2 = \frac{i_X}{2} = \frac{i_1}{2} = -2 \text{ mA}$$

To find the voltage drop across the dependent source, we use a voltage loop on the right side.