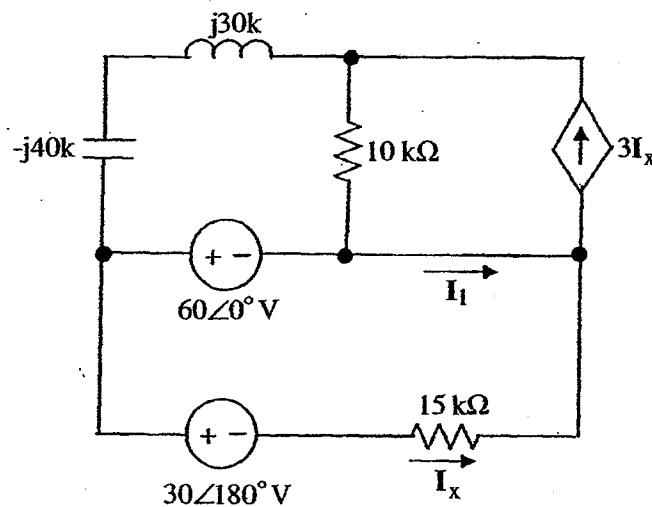


HW #10 Cont.

4.



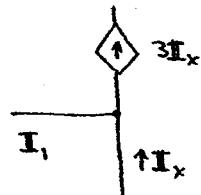
- A frequency-domain circuit is shown above. Write the value of phasor  $I_1$  in polar form.
- Given  $\omega = 53.13 \text{ rad/s}$ , write a numerical time-domain expression for  $i_1(t)$ , the inverse phasor of  $I_1$ .

HW #10 Cont.

Su 05

ECE 1000

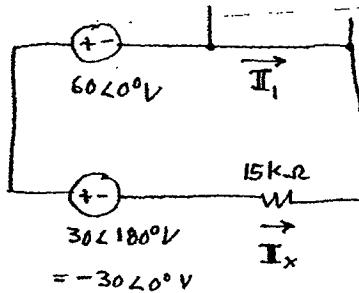
sol'n: 4. a) Sum of currents for node on right side:



$$\text{we see that } I_1 = 2I_x$$

from sum of currents  
out of node = 0.

From the bottom half of the circuit, we  
can compute  $I_x$  directly:



From v-loop we have

$$I_x = \frac{60<0^\circ V - 30<0^\circ V}{15k\Omega}$$

$$I_x = \frac{30<0^\circ V}{15k\Omega}$$

$$I_x = 6 \text{ mA} <0^\circ$$

$$\text{So, } I_1 = 2I_x = 12 \text{ mA} <0^\circ$$

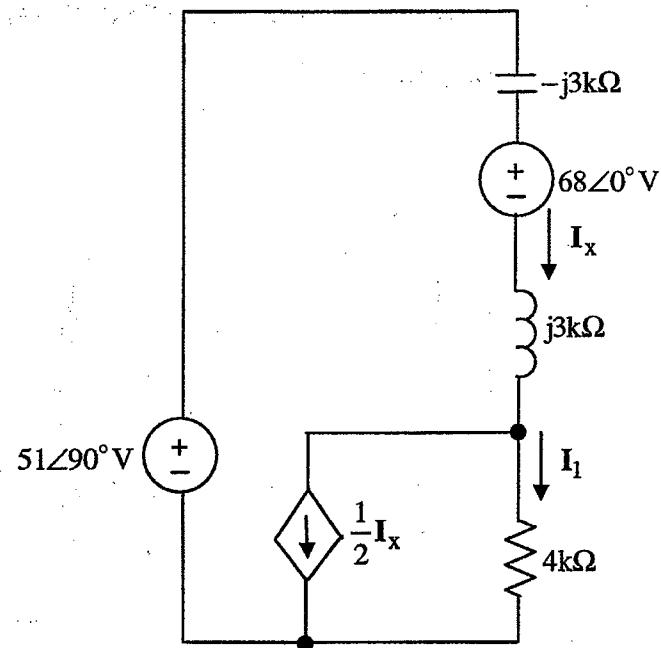
$$I_1 = 12 <0^\circ \text{ mA}$$

b)

$$i_1(t) = 12 \cos(53.13t) \text{ mA}$$

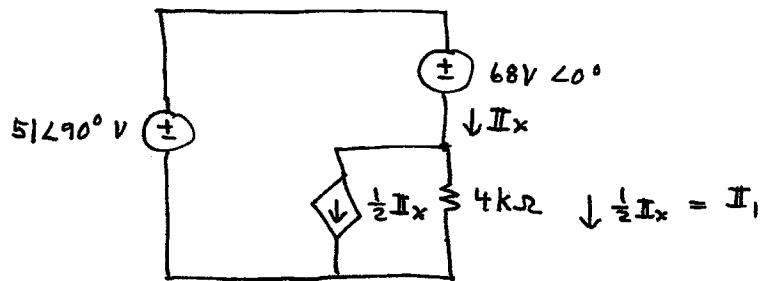
or  $\pi t$

3.



- A frequency-domain circuit is shown above. Write the value of phasor  $I_1$  in polar form.
- Given  $\omega = \pi \text{ rad/s}$ , write a numerical time-domain expression for  $i_1(t)$ , the inverse phasor of  $I_1$ .

Sol'n: 3.a) The  $-j3k\Omega$  and  $j3k\Omega$  sum to zero and act like a wire. Thus, they do not affect  $\mathbb{I}_x$ . So we have:



Clearly,  $\frac{1}{2} \mathbb{I}_x$  flows thru the  $4k\Omega$  (for sum of currents at node above  $4k\Omega = 0$ ).

But the current thru  $4k\Omega$  is  $\frac{51\angle 90^\circ V - 68\angle 0^\circ V}{4k\Omega}$

$$\text{or } \frac{1}{2} \mathbb{I}_x = \mathbb{I}_1 = \frac{17 \cdot 3\angle 90^\circ - 17 \cdot 4\angle 0^\circ V}{4k\Omega}$$

$$\mathbb{I}_1 = 17 \frac{j3 - 4}{4k\Omega} = \frac{17}{4} (-4 + j3) = \frac{17.5}{4} V \angle 143^\circ \text{ mA}$$

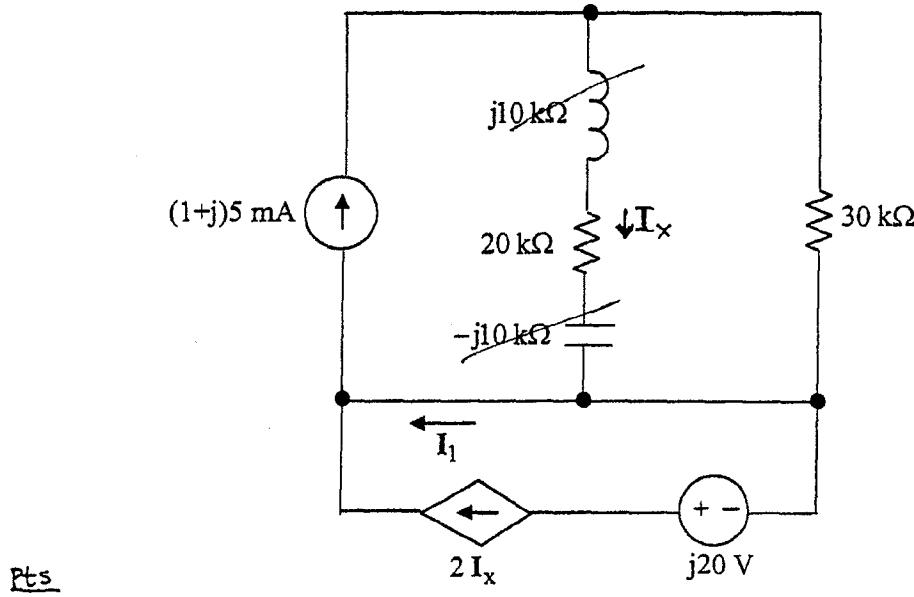
$$\boxed{\mathbb{I}_1 = -17 + j 12.75 \text{ mA} = 21.25 \angle 143^\circ \text{ mA}}$$

b)

$$i_1(t) = 17 \cos(\pi t + 180^\circ) - 12.75 \sin(\pi t) \text{ mA}$$

$$" = 21.25 \cos(\pi t + 143^\circ) \text{ mA}$$

4. (25 points)



Pts

- 20 pts a. A frequency-domain circuit is shown above. Write the value of  $I_1$  in polar form.
- 5 pts b. Given  $\omega = 100 \text{ k rad/s}$ , write a numerical time-domain expression for  $i_1(t)$ , the inverse phasor of  $I_1$ .

Sol'n: 4.a)  $j10\text{k}\Omega - j10\text{k}\Omega = 0\Omega$  so L and C cancel.

Current divider for  $20\text{k}\Omega$  and  $30\text{k}\Omega$ .

$$\therefore I_x = (1+j) 5 \text{ mA} \cdot \frac{30\text{k}\Omega}{20\text{k}\Omega + 30\text{k}\Omega} = (1+j) 3 \text{ mA}$$

Find  $I_1$  from sum of currents at node  
on left side:

$$(1+j) 5 \text{ mA} - 2 \underbrace{(1+j) 3 \text{ mA}}_{I_x} - I_1 = 0$$

$$I_1 = (1+j) 5 \text{ mA} - 2 (1+j) 3 \text{ mA} = (1+j)(5-6) \text{ mA}$$

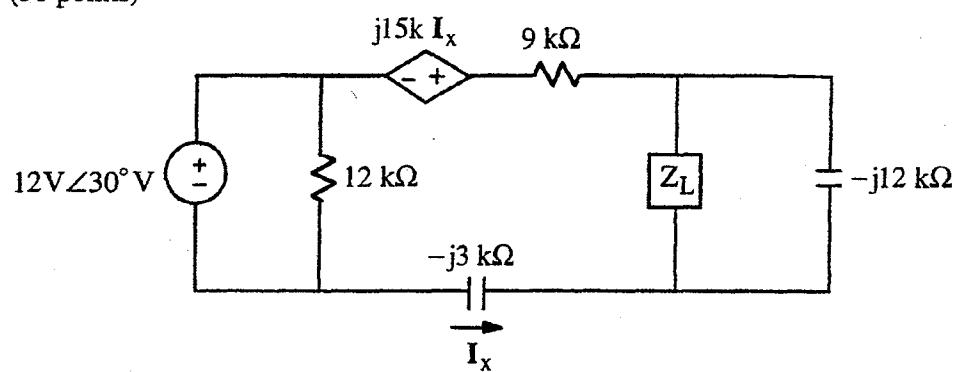
$$I_1 = (1+j)(-1) \text{ mA} = -(1+j) \text{ mA}$$

$I_1 = -\sqrt{2} \angle 45^\circ \text{ mA}$	$= \sqrt{2} \angle -135^\circ \text{ mA}$
--	---

or  $225^\circ$

b)  $i_1(t) = \sqrt{2} \cos(100kt - 135^\circ) \text{ mA}$

3. (30 points)



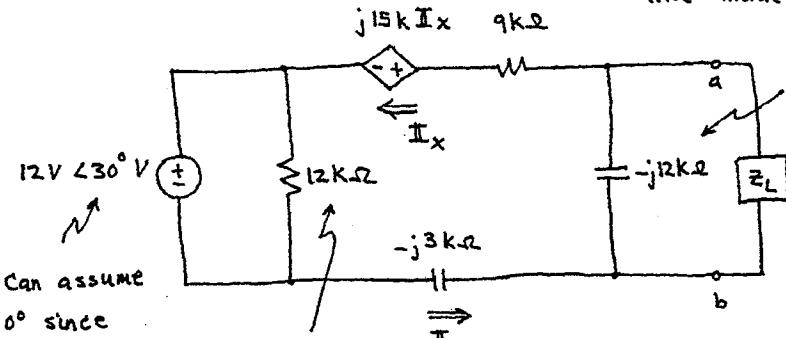
Pts

- 15 a. Choose the value of  $Z_L$  that will absorb maximum average power.  
15 b. Calculate the value of that maximum average power absorbed by  $Z_L$ .

(46)

solt'n 3. a)  $z_L = z_{Th}^*$  for max power xfer

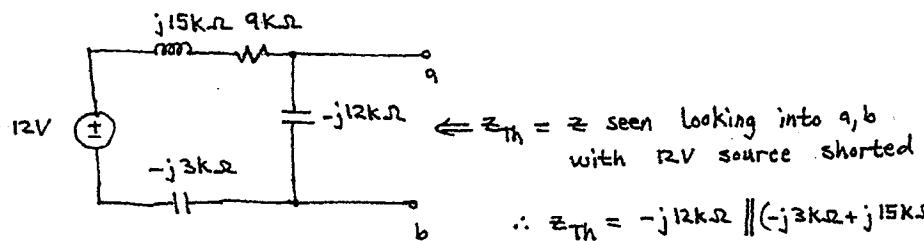
Since  $I_x$  flows thru it, this looks like inductor with  $z = j15k\Omega$ .



Can assume  
0° since  
phase angle  
doesn't affect  
are pwr  
not seen  
in Thvenin  
equivalent since  
it is across V source

We can put C on left  
side of  $z_L$  since  
they are in  
parallel

Circuit for determining Thvenin equivalent:



$\leftarrow z_{Th} = z \text{ seen looking into } a, b \text{ with } 12V \text{ source shorted}$

$$\therefore z_{Th} = -j12k\Omega \parallel (-j3k\Omega + j15k\Omega + 9k\Omega)$$

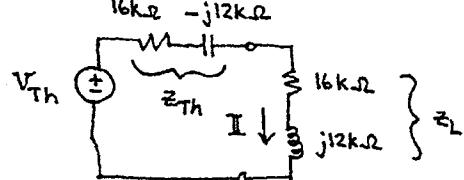
$$" = -j12k\Omega \parallel (9k\Omega + j12k\Omega)$$

$$" = \frac{-j12k\Omega (9k\Omega + j12k\Omega)}{-j12k\Omega + 9k\Omega + j12k\Omega}$$

$$" = -j \frac{12k\Omega (9k\Omega + j12k\Omega)}{9k\Omega}$$

$$" = -j12k\Omega + 16k\Omega$$

b)



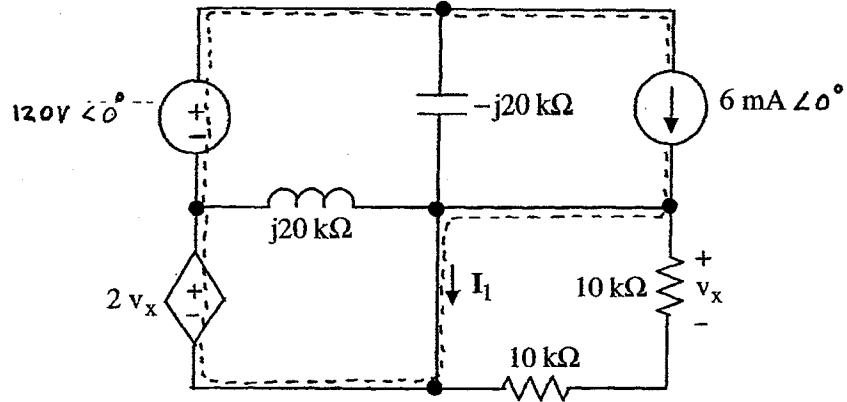
$$P \equiv (\text{ave pwr in } z_L) = \frac{|I|^2}{2} R \text{ where } R = 16k\Omega \text{ in } z_L$$

$$I = \frac{V_{Th}}{16k\Omega - j12k\Omega + 16k\Omega + j12k\Omega} = \frac{V_{Th}}{32k\Omega}$$

$$V_{Th} = 12V \cdot \frac{-j12k\Omega}{-j12k\Omega + -j3k\Omega + j15k\Omega + 9k\Omega} = 12V \left( \frac{-j12k\Omega}{9k\Omega} \right) = -j16V$$

$$\therefore P = \frac{\left| \frac{-j16V}{32k\Omega} \right|^2 \cdot 16k\Omega}{2} = \frac{|-j1|^2 V^2}{|32k\Omega|^2} \cdot \frac{16k\Omega}{2} = \frac{1V^2 \cdot 8}{4k\Omega} \text{ or } P = 2 \text{ mW}$$

4. (25 points)



Pts

- 20 a. A frequency-domain circuit is shown above. Write the value of phasor  $\mathbf{I}_1$  in polar form.
- 5 b. Given  $\omega = \pi$  rad/s, write a numerical time-domain expression for  $i_1(t)$ , the inverse phasor of  $\mathbf{I}_1$ .

Sol'n: a) Since the two  $10\text{k}\Omega$  resistors are shorted by wires.

$\therefore$  There is no  $v$  drop across the  $10\text{k}\Omega$  resistors,  
and  $v_x = 0\text{V}$ .

Thus, the  $2v_x$  dependent source =  $0\text{V}$  = wire  
Superposition Case I: 6mA on, 120V off = wire.

It follows that all of the 6mA from the independent current source flows in the wires (shown as dashed lines above).

$$\therefore \mathbf{I}_{11} = 6\text{mA} < 0^\circ$$

Case II: 120V on, 6mA off = open circuit  
we observe that the  $-j20\text{k}\Omega$  is directly across the 120V source, given the wires shown as dashed lines.

$$\therefore \mathbf{I}_{12} = \frac{120\text{V} < 0^\circ}{-j20\text{k}\Omega} = j6\text{mA} = 6\text{mA} < 90^\circ$$

$$\text{Thus, } \mathbf{I}_1 = \mathbf{I}_{11} + \mathbf{I}_{12} = 6\text{mA} \cdot (1+j)$$

or 
$$\boxed{\mathbf{I}_1 = \sqrt{2} \cdot 6\text{mA} < 45^\circ}$$

b) 
$$\boxed{i_1(t) = \sqrt{2} \cdot 6\text{mA} \cos(\pi t + 45^\circ)}$$