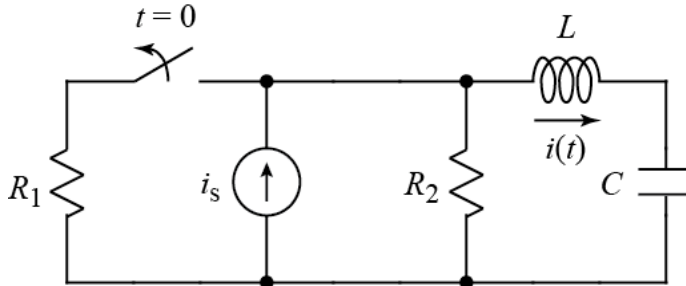


Ex:



After being closed for a long time, the switch opens at  $t = 0$ .

$$L = 1 \mu\text{H} \quad C = 16 \mu\text{F} \quad R_1 = 0.15 \Omega \quad R_2 = 0.1 \Omega$$

If  $i_s = 30 \text{ mA}$ , find  $i(t)$  for  $t > 0$ .

**SOL'N:** We may perform the following initial steps in any order:

- 1) Find characteristic roots for the parallel RLC circuit (for  $t > 0$ )
- 2) Find the final value of  $i(t)$  as  $t \rightarrow \infty$ , which is the  $A_3$  (constant) term in the solution.
- 3) Find the initial values of energy variables:  $i_L(0^+) = i_L(0^-)$  and  $v_C(0^+) = v_C(0^-)$

We will perform the steps in the order listed. First, we find the characteristic roots.

$$s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2}$$

After time  $t = 0$ ,  $R_1$  will be disconnected from the circuit and may be ignored. The characteristic roots are unaffected by source  $i_s$ . We have a series RLC circuit for which the value of  $\alpha$  is one-half the inverse L/R time constant:

$$\alpha = \frac{R_2}{2L} = \frac{0.1 \Omega}{2 \cdot (1\mu\text{H})} = 50 \text{ k/s}$$

For both parallel and series RLC circuits, the resonant frequency,  $\omega_0$ , is the inverse of the square root of the product of L and C:

$$\omega_0^2 = \frac{1}{LC} = \frac{1}{1 \mu\text{H} \cdot 16 \mu\text{F}} = (250\text{k})^2 (\text{r/s})^2$$

Substituting into the equation for the roots yields the following:

$$s_{1,2} = -50 \text{ kr/s} \pm \sqrt{(50\text{k})^2 - (250\text{k})^2} \text{ r/s} \approx -50 \text{ kr/s} \pm j245 \text{ kr/s}$$

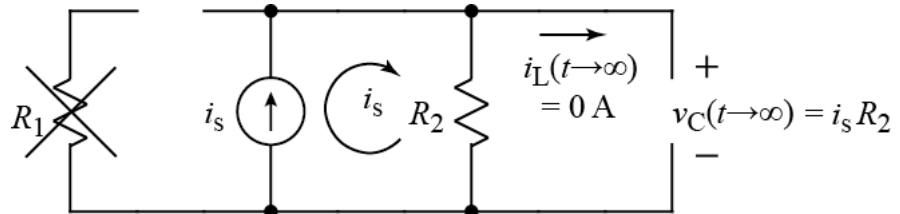
The imaginary part of the roots is the damping frequency, which we may compute by taking the negative of the quantity inside the square root:

$$\omega_d = \sqrt{\omega_0^2 - \alpha^2} \approx 245 \text{ kr/s}$$

For complex roots, we use the following form of solution:

$$v(t) = A_1 e^{-\alpha t} \cos(\omega_d t) + A_2 e^{-\alpha t} \sin(\omega_d t) + A_3$$

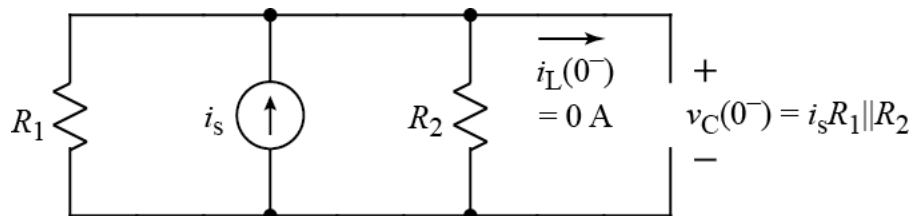
Now we proceed to step 2, where we find the  $A_3$  value from  $v(t)$  as  $t \rightarrow \infty$ . (The exponential terms decay as  $t \rightarrow \infty$ , leaving only  $A_3$ .) We assume the circuit values become constant as  $t \rightarrow \infty$ , causing the  $L$  to act like a wire and the  $C$  to act like an open circuit. The switch is also open, detaching  $R_1$  on the left, as shown in the diagram below.



Since the  $C$  acts like an open circuit, the current in the  $L$  is zero, and we have  $A_3 = i(t \rightarrow \infty) = 0 \text{ A}$ .

Third, we find the initial values of energy variables:  $i_L(0^+) = i_L(0^-)$  and  $v_C(0^+) = v_C(0^-)$ . At  $t = 0^-$ , we assume circuit values are constant, causing the  $L$  to act like a wire and the  $C$  to act like an open circuit. The switch is closed, connecting the parallel RLC to  $R_1$  on the left.

$t = 0^-$ :



The  $C$ , acting like an open, makes the initial  $L$  current zero:

$$i_L(0^-) = 0 \text{ A}$$

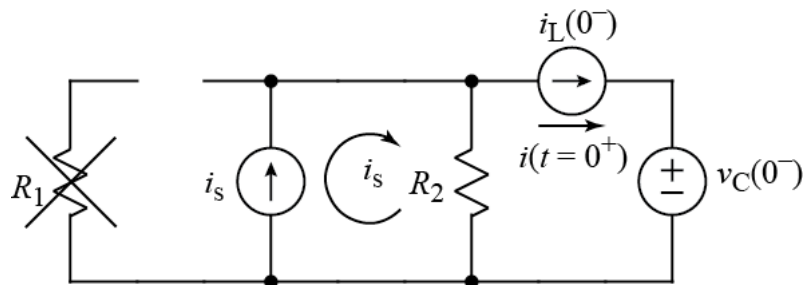
The two  $R$ 's are in parallel across the voltage source. We may use Ohm's law to find the voltage across the rails. This equals the voltage across the  $C$ :

$$v_C(0^+) = v_C(0^-) = i_s \cdot R_1 \parallel R_2 = 30 \text{ mA} \cdot 0.1 \parallel 0.15 \Omega$$

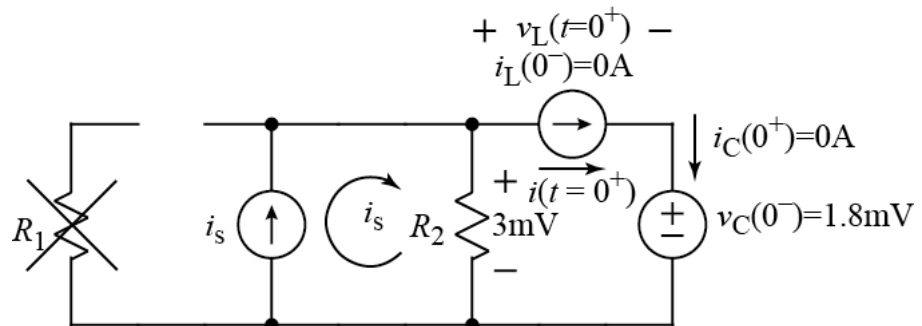
or

$$v_C(0^+) = v_C(0^-) = 30 \text{ mA} \cdot 0.06 \Omega = 1.8 \text{ mV}$$

At time  $t = 0^+$ , we treat the  $C$  as a voltage source of 1.8 mV and the  $L$  as a current source of 0 A. (The switch is also open, removing  $R_1$  from the circuit.)



Because the current in the  $L$  is zero, all of  $i_s$  flows through  $R_2$ . The voltage across  $R_2$ , by Ohm's law, is  $i_s R_2 = 30 \text{ mA} \cdot 0.1 \Omega = 3 \text{ mV}$ .



We can solve the circuit for any voltage or current at  $t = 0^+$ . Using Kirchhoff's laws for current sums at nodes, we have a voltage loop on the right that yields the voltage across the  $L$  at  $t = 0^+$ :

$$v_L(0^+) = 3 \text{ mV} - 1.8 \text{ mV} = 1.2 \text{ mV}$$

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For the  $C$ , no current flows in the  $L$ , and so no current flows in the  $C$ :

$$i_C(0^+) = 0 \text{ A}$$

Now we are ready to find  $A_1$  and  $A_2$  by matching our symbolic solutions

$$i(0^+) = A_1 + A_3 = A_1$$

and

$$\left. \frac{di(t)}{dt} \right|_{t=0^+} = -\alpha A_1 + \omega_d A_2$$

to circuit values for these quantities. We have already found  $i(0^+)$  to be 0 A. Matching this to the symbolic solution for  $t = 0^+$ , we have the following:

$$A_1 = 0 \text{ V}$$

For the value of the derivative in the circuit, we first write our variable, ( $i(t)$  in this case), in terms of energy (or state) variables,  $i_L$  and/or  $v_C$ . Here,  $i$  is  $i_L$ :

$$i(t) = i_L(t)$$

Next, we differentiate, and use the component equations involving  $d/dt$  for  $L$  and/or  $C$ :

$$\left. \frac{d}{dt} i(t) \right|_{t=0^+} = \left. \frac{d}{dt} i_L(t) \right|_{t=0^+} = \frac{v_L(0^+)}{L} = \frac{1.2 \text{ mV}}{1 \mu\text{H}} = 1.2 \text{ kA/s}$$

We equate this to the symbolic derivative:

$$\left. \frac{d}{dt} i(t) \right|_{t=0^+} = -\alpha A_1 + \omega_d A_2 = \omega_d A_2 = 1.2 \text{ kA/s}$$

or

$$A_2 = \frac{1.2 \text{ kA/s}}{\omega_d} \approx \frac{1.2 \text{ kA/s}}{245 \text{ kr/s}} \approx 4.9 \text{ mA/s}$$

Plugging in values gives the solution for  $v(t > 0)$ :

$$i(t) = 4.9 e^{-50 \text{ k/s} \cdot t} \sin(245 \text{ kr/s} \cdot t) \text{ mA}$$