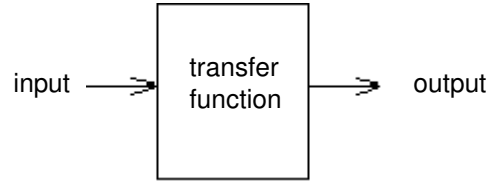


Yesterday we drew a block diagram on the board. Let's examine those block a little more closely

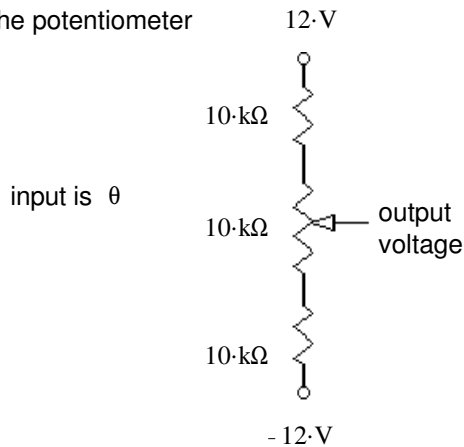


What's inside?

How are the input and output related? If you know the input, how do you find the output? Sometimes we can just multiply the input by the expression in the box to get the output. Then the expression in the box is called a transfer function.

In that case, the transfer function =  $\frac{\text{output}}{\text{input}}$

A very simple case, the potentiometer

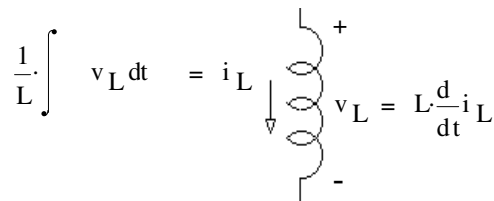
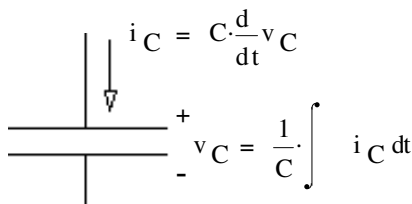


We measure the voltage over its range of motion to find  $K_p$ .

$$K_p = \frac{\text{output}}{\text{input}} = \frac{4 \cdot V - -4 \cdot V}{270 \cdot \text{deg}} = 29.63 \cdot \frac{\text{mV}}{\text{deg}} = 1.7 \cdot \frac{V}{\text{rad}}$$

"Zero" must be in the center

Nice... too bad it works for so few things in the time domain! Simple voltage dividers, amplifiers, and ?? In electrical systems there are always capacitance and inductance.



We'll have to avoid capacitors and inductors-- they're too complicated... You can't just multiply when there are differentials involved

How about the mechanical world?  $F = ma$ , Great, no differentials... uh, except...  $F = m \cdot a = m \cdot \frac{d}{dt} v = m \cdot \frac{d^2}{dt^2} x$

And then there are springs:  $F = k \cdot x = k \cdot \int v dt = k \cdot \int \int a dt dt$

Isn't there some way that we could possibly replace all this differentiation and integration with multiplication and division?

Laplace transforms

$\frac{d}{dt}$  operation can be replaced with  $s$ , and  $\int dt$  can be replaced by  $\frac{1}{s}$

## ECE 2210 Lecture 2 notes p2

Recall from your Ordinary Differential Equations class, the Laplace transform method of solving differential equations. The Laplace transform allowed you to change time-domain functions to frequency-domain functions.

1) Transform your signals into the frequency domain with the Laplace transform.

$$F(s) = \int_0^{\infty} f(t) \cdot e^{-st} dt \quad \text{Unilateral Laplace transform}$$

2) Solve your differential equations with plain old algebra, where:

$$\frac{d}{dt} \text{ operation can be replaced with } s, \quad \text{and} \quad \int \blacksquare dt \text{ can be replaced by } \frac{\blacksquare}{s}$$

3) Transform your result back to the time domain with the inverse Laplace transform.

$$f(t) = \frac{1}{2\pi j} \int_{c-j\infty}^{c+j\infty} F(s) \cdot e^{st} ds \quad \text{OK, truth be told, we never actually use the inverse Laplace transform. We use tables instead.}$$

So, the first step is to transform the signals into the frequency domain with the Laplace transform. Maybe we ought to talk a little about signals first...

### Signals

For us: A time-varying voltage or current that carries information.

In some unpredictable fashion	Audio, video, position, temperature, digital data, etc...

DC is not a signal, Neither is a pure sine wave. If you can predict it, what information is it providing??  
Neither DC nor pure sine wave have any "bandwidth".

Recall Fourier series: Any periodic waveform can be represented by a series of sinewaves of different frequencies.

### Laplace transforms

Let's evaluate some of these and see if we can make a table

**Ex. 1**  $f(t) = \delta(t)$  The Impulse or "Dirac" function, not a very likely signal in real life.

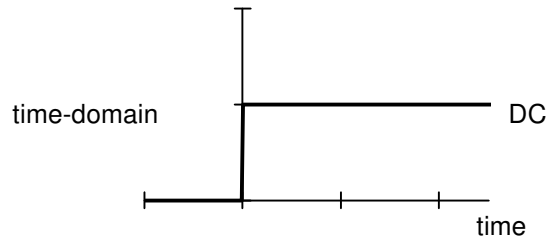
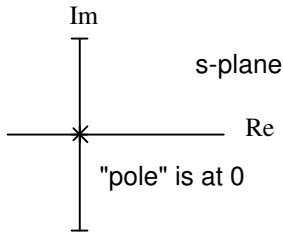
$$\begin{aligned}
 F(s) &= \int_0^{\infty} \delta(t) \cdot e^{-st} dt & \text{but: } \delta(t) \cdot g(t) &= \delta(t) \cdot g(0) & \text{so:} \\
 & & \text{any function} & & \\
 &= \int_0^{\infty} \delta(t) \cdot e^{-s \cdot 0} dt & &= \int_0^{\infty} \delta(t) \cdot 1 dt &= 1
 \end{aligned}$$

**Ex. 2**  $f(t) = u(t)$  The unit-step function, a constant value (DC) signal

$$F(s) = \int_0^{\infty} u(t) \cdot e^{-st} dt \quad \text{but: } u(t) = 1 \text{ from } 0 \text{ to } \infty$$

$$= \int_0^{\infty} 1 \cdot e^{-st} dt = \left. \frac{1}{-s} \cdot e^{-st} \right|_0^{\infty} = \frac{1}{-s} \cdot e^{-s \cdot \infty} - \frac{1}{-s} \cdot e^{-s \cdot 0} = 0 - \frac{1}{-s} \cdot (1) = \frac{1}{s}$$

if  $s > 0$



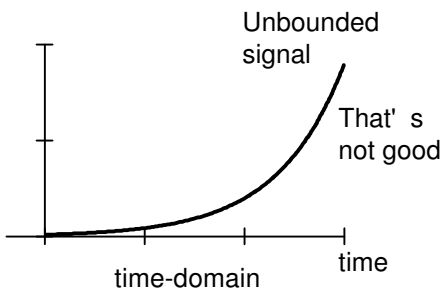
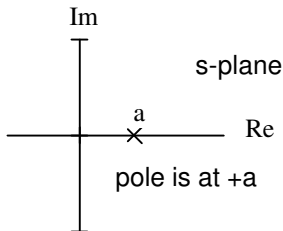
**Ex. 3**  $f(t) = u(t) \cdot e^{at}$

$$F(s) = \int_0^{\infty} e^{at} \cdot e^{-st} dt = \int_0^{\infty} e^{(a-s)t} dt = \left. \frac{1}{(a-s)} \cdot e^{(a-s)t} \right|_0^{\infty}$$

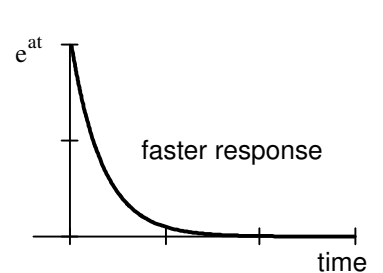
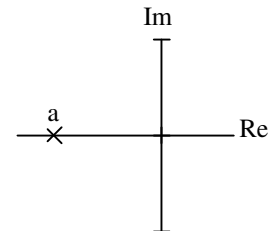
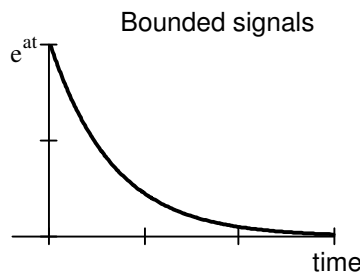
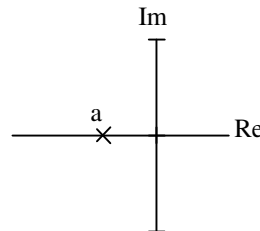
$$= \frac{1}{(a-s)} \cdot e^{(a-s) \cdot \infty} - \frac{1}{(a-s)} \cdot e^{(a-s) \cdot 0} = 0 - \frac{1}{(a-s)} \cdot (1) = \frac{1}{s-a}$$

if  $s > a$  "pole" is at +a

for positive a values



for negative a values



This is the single most-important Laplace transform case. In fact we really don't need any others. Ex.1 can be thought of as this case with  $a = -\infty$ . Ex. 2 can be thought of as  $a = 0$ . And finally, all sinusoids can be made from exponentials if you let the poles ( $a$ ) be complex. Remember Euler's equations...

Euler's equations  $e^{j\omega t} = \cos(\omega t) + j \cdot \sin(\omega t)$

$e^{(\alpha + j\omega)t} = e^{\alpha t} \cdot (\cos(\omega t) + j \cdot \sin(\omega t))$

**Pole Location(s) correspond to the type of signal.**

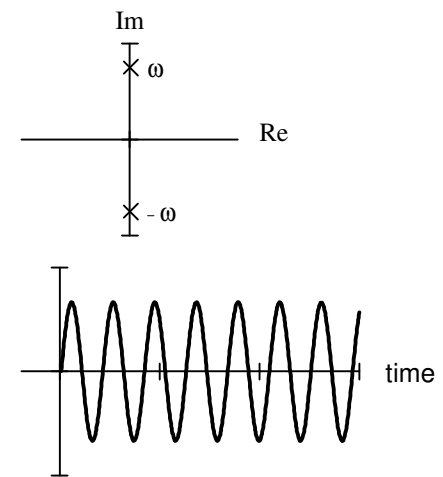
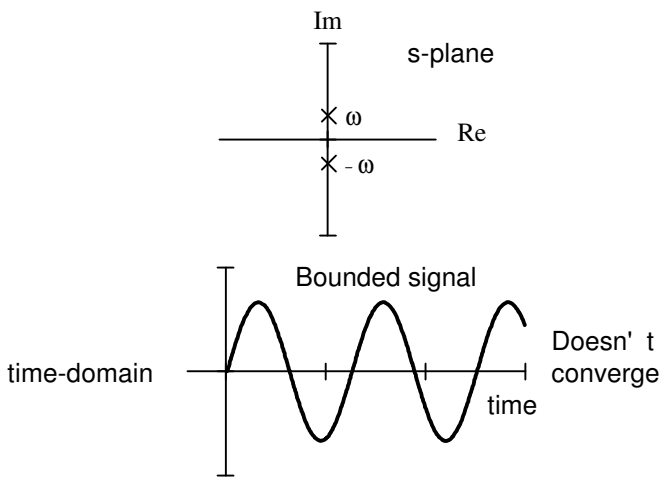
# ECE 3510 Lecture 2 notes p4

Euler's equations  $\cos(\omega \cdot t) = \frac{e^{j\omega t} + e^{-j\omega t}}{2}$

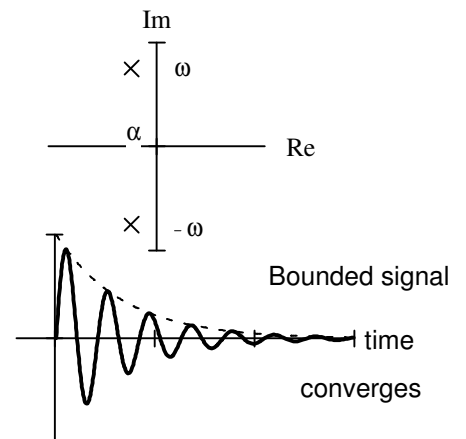
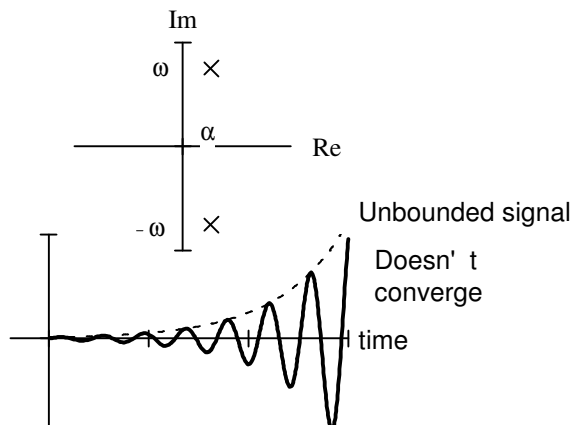
$\sin(\omega \cdot t) = \frac{e^{j\omega t} - e^{-j\omega t}}{2 \cdot j}$

**Ex. 4**  $f(t) = u(t) \cdot \cos(\omega \cdot t)$

$$\begin{aligned}
 F(s) &= \int_0^{\infty} \cos(\omega \cdot t) \cdot e^{-s \cdot t} dt = \int_0^{\infty} \left( \frac{e^{j\omega t} + e^{-j\omega t}}{2} \right) \cdot e^{-s \cdot t} dt = \int_0^{\infty} \frac{e^{(j\omega - s) \cdot t} + e^{-(j\omega + s) \cdot t}}{2} dt \\
 &= \frac{1}{2} \int_0^{\infty} e^{(j\omega - s) \cdot t} dt + \frac{1}{2} \int_0^{\infty} e^{-(j\omega + s) \cdot t} dt \\
 &= \frac{1}{2} \cdot \left[ \frac{1}{j\omega - s} \right] \cdot e^{(j\omega - s) \cdot t} \Bigg|_0^{\infty} + \frac{1}{2} \cdot \left[ \frac{1}{-(j\omega + s)} \right] \cdot e^{-(j\omega + s) \cdot t} \Bigg|_0^{\infty} \\
 &= 0 - \frac{1}{2} \cdot \left[ \frac{1}{j\omega - s} \right] \cdot (1) + 0 - \frac{1}{2} \cdot \left[ \frac{1}{-(j\omega + s)} \right] \cdot (1) = \frac{-1}{-2 \cdot j\omega - 2 \cdot s} + \frac{-1}{2 \cdot j\omega - 2 \cdot s} \\
 &= \frac{1}{2 \cdot j\omega + 2 \cdot s} + \frac{-1}{2 \cdot j\omega - 2 \cdot s} = \frac{(2 \cdot j\omega - 2 \cdot s) - (2 \cdot j\omega + 2 \cdot s)}{(2 \cdot j\omega + 2 \cdot s) \cdot (2 \cdot j\omega - 2 \cdot s)} \\
 &= \frac{-4 \cdot s}{(2 \cdot j\omega + 2 \cdot s) \cdot (2 \cdot j\omega - 2 \cdot s)} = \frac{-4 \cdot s}{4 \cdot j^2 \cdot \omega^2 - 4 \cdot s^2} = \frac{-s}{-\omega^2 - s^2} = \frac{s}{\omega^2 + s^2}
 \end{aligned}$$



What if the poles have a real component?  $f(t) = u(t) \cdot e^{\sigma t} \cdot \sin(\omega \cdot t)$



**Ex. 5**  $f(t) = u(t) \cdot t \cdot e^{at}$        $F(s) = \int_0^{\infty} t \cdot e^{at} \cdot e^{-st} dt = \int_0^{\infty} t \cdot e^{(a-s)t} dt$

Remember integration by parts:

$$\int f(t) \cdot \frac{d}{dt} g(t) dt = f(t) \cdot g(t) - \int g(t) \cdot \frac{d}{dt} f(t) dt$$

choose:  $f(t) = t$       from which:  $\frac{d}{dt} f(t) = 1$

and:  $\frac{d}{dt} g(t) = e^{(a-s)t}$       from which:  $g(t) = \int e^{(a-s)t} dt = \frac{e^{(a-s)t}}{(a-s)}$

$$\begin{aligned} F(s) &= \int_0^{\infty} t \cdot e^{(a-s)t} dt = t \cdot \frac{e^{(a-s)t}}{(a-s)} \Big|_0^{\infty} - \int_0^{\infty} \frac{e^{(a-s)t}}{(a-s)} \cdot (1) dt \\ &= t \cdot \frac{e^{(a-s)t}}{(a-s)} \Big|_0^{\infty} - \frac{e^{(a-s)t}}{(a-s)^2} \Big|_0^{\infty} \\ &= 0 - 0 - \left[ 0 - \frac{1}{(a-s)^2} \right] \\ &= \frac{1}{(a-s)^2} = \frac{1}{(s-a)^2} \end{aligned}$$

The easy way:

Use the "multiplication by time" property # 5 on p.8 of the textbook

$$t \cdot x(t) \iff -\frac{d}{ds} X(s)$$

$$t \cdot e^{at} \iff -\frac{d}{ds} \left( \frac{1}{s-a} \right) = -\frac{d}{ds} [(s-a)^{-1}] = -\frac{1}{-1} \cdot \frac{1}{(s-a)^2} \cdot \left[ \frac{d}{ds} (s-a) \right] = \frac{1}{(s-a)^2} \cdot 1 = \frac{1}{(s-a)^2}$$

