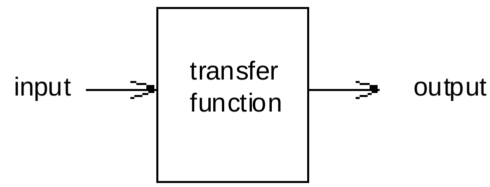
ECE 3510 Lecture 2 notes Introduction to Feedback Systems

A. Stolp
1/8/08,
1/25/09

Yesterday we drew a block diagram on the board.
Let's examine those block a little more closely



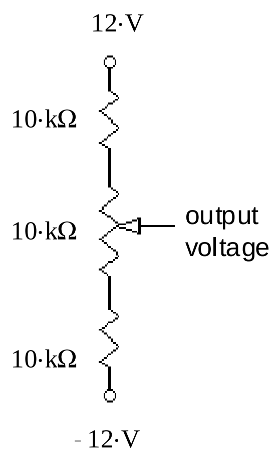
What's inside?

How are the input and output related? If you know the input, how do you find the outout? Sometimes we can just multiply the input by the expression in the box to get the output. Then the expresion in the box is called a transfer function.

In that case, the transfer function = $\dfrac{\text{output}}{\text{input}}$

A very simple case, the potentiometer

input is θ



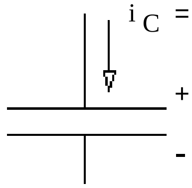
We measure the voltage over its range of motion to find K_p.

$$K_p = \frac{\text{output}}{\text{input}} = \frac{4\cdot V - -4\cdot V}{270\cdot deg} = 29.63 \cdot \frac{mV}{deg} = 1.7\cdot\frac{V}{rad}$$

range of motion

"Zero" must be in the center

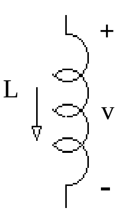
Nice... too bad it works for so few things in the time domain! Simple voltage dividers, amplifiers, and ?? In electrical systems there are always capacititance and inductance.



$$i_C = C\cdot\frac{d}{dt}v_C$$

$$v_C = \frac{1}{C}\cdot\int i_C\,dt$$

$$\frac{1}{L}\cdot\int v_L\,dt = i_L$$



$$v_L = L\cdot\frac{d}{dt}i_L$$

We'll have to avoid capacitors and inductors-- they're too complicated... You can't just multiply when there are differentials involved

How about the mechanical world? $F = ma$, Great, no differentials... uh, except... $F = m\cdot a = m\cdot\frac{d}{dt}v = m\cdot\frac{d^2}{dt^2}x$

And then there are springs: $F = k\cdot x = k\cdot\int v\,dt = k\cdot\int\int a\,dt\,dt$

Isn't there some way that we could possibly replace all this differentiation and integration with multiplication and division?

Laplace transforms

$\frac{d}{dt}$ operation can be replaced with s, and $\int \blacksquare\,dt$ can be replaced by $\frac{1}{s}$

Recall from your Ordinary Differential Equations class, the Laplace transform method of solving differential equations. The Laplace transform allowed you to change time-domain functions to frequency-domain functions.

1) Transform your signals into the frequency domain with the Laplace transform.

$$F(s) = \int_0^{\infty} f(t)\cdot e^{-s\cdot t}\,dt$$ Unilateral Laplace transform

2) Solve your differential equations with plain old algebra, where:

$\frac{d}{dt}$ operation can be replaced with s, and $\int \blacksquare\,dt$ can be replaced by $\frac{1}{s}$

3) Transform your result back to the time domain with the inverse Laplace transform.

$$f(t) = \frac{1}{2\cdot\pi\cdot j}\cdot\int_{c-j\infty}^{c+j\infty} F(s)\cdot e^{s\cdot t}\,ds$$

OK, truth be told, we never actually use the inverse Laplace transform. We use tables instead.

So, the first step is to transform the signals into the frequency domain with the Laplace transform.
Maybe we ought to talk a little about signals first...

Signals

For us: A time-varying voltage or current that carriers information.

| (time-varying) → In some unpredictable fashion
| (information) → Audio, video, position, temperature, digital data, etc...

DC is not a signal, Neither is a pure sine wave. If you can predict it, what information is it providing??
Neither DC nor pure sine wave have any "bandwidth".

Recall Fourier series: Any periodic waveform can be represented by a series of sinewaves of different frequencies.

Laplace transforms

Let's evaluate some of these and see if we can make a table

Ex. 1 $f(t) = \delta(t)$ The Impulse or "Dirac" function, not a very likely signal in real life.

$$F(s) = \int_0^{\infty} \delta(t)\cdot e^{-s\cdot t}\,dt$$

but: $\delta(t)\cdot g(t) = \delta(t)\cdot g(0)$ (any function) so:

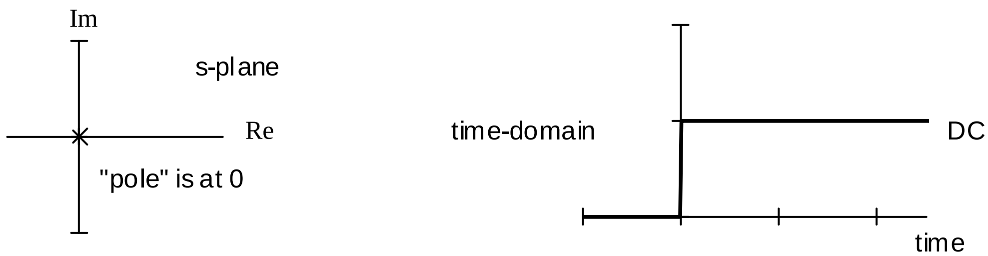
$$= \int_0^{\infty} \delta(t)\cdot e^{-s\cdot 0}\,dt = \int_0^{\infty} \delta(t)\cdot 1\,dt = 1$$

Ex. 2 $f(t) = u(t)$ The unit-step function, a constant value (DC) signal

$$F(s) = \int_0^\infty u(t)\cdot e^{-s\cdot t}\,dt \qquad \text{but:} \quad u(t) = 1 \text{ from } 0 \text{ to } \infty$$

$$= \int_0^\infty 1\cdot e^{-s\cdot t}\,dt = \frac{1}{-s}\cdot e^{-s\cdot t}\Big|_0^\infty = \frac{1}{-s}\cdot \overset{0}{e^{-s\cdot\infty}} - \frac{1}{-s}\cdot \overset{1}{e^{-s\cdot 0}} = 0 - \frac{1}{-s}\cdot(1) = \frac{1}{s}$$

if $s > 0$

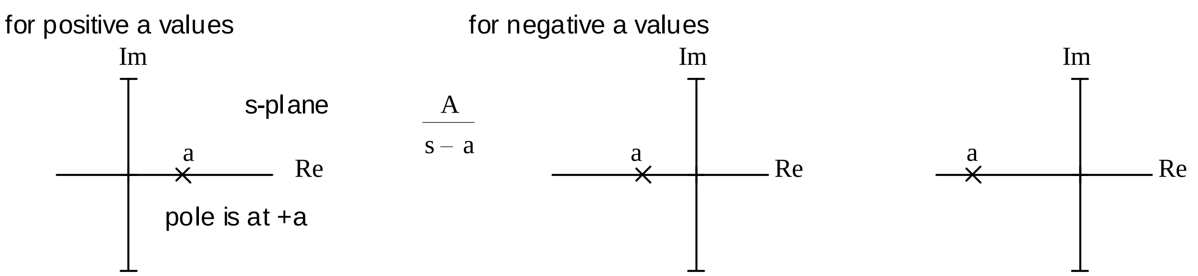


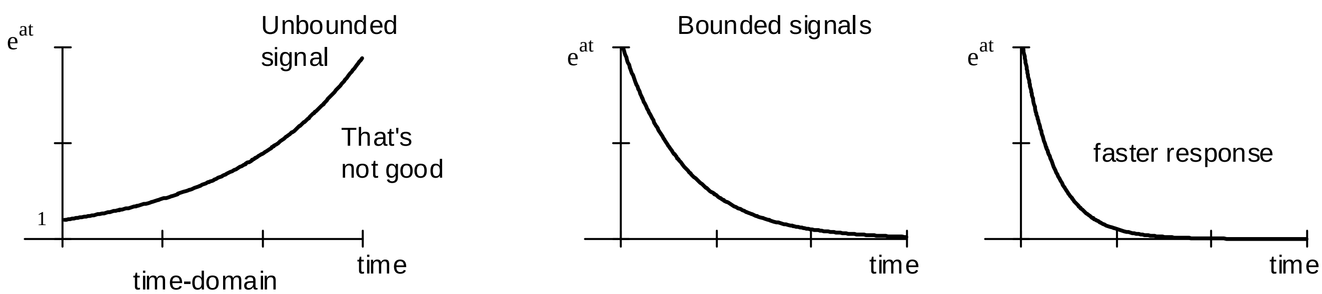
Ex. 3 $f(t) = u(t)\cdot e^{at}$

$$F(s) = \int_0^\infty e^{at}\cdot e^{-s\cdot t}\,dt = \int_0^\infty e^{(a-s)\cdot t}\,dt = \frac{1}{(a-s)}\cdot e^{(a-s)\cdot t}\Big|_0^\infty$$

$$= \frac{1}{(a-s)}\cdot \overset{0}{e^{(a-s)\cdot\infty}} - \frac{1}{(a-s)}\cdot \overset{1}{e^{(a-s)\cdot 0}} = 0 - \frac{1}{(a-s)}\cdot(1) = \frac{1}{s-a}$$

if $s > a$ "pole" is at +a





This is the single most-important Laplace transform case. In fact we really don't need any others. Ex.1 can be thought of as this case with $a = -\infty$. Ex. 2 can be thought of as $a = 0$. And finally, all sinusoids can be made from exponentials if you let the poles (a) be complex. Remember Euler's equations...

Euler's equations $e^{j\cdot\omega\cdot t} = \cos(\omega t) + j\cdot\sin(\omega t)$ $\qquad e^{(\alpha\cdot t + j\cdot\omega\cdot t)} = e^{\alpha\cdot t}\cdot(\cos(\omega t) + j\cdot\sin(\omega t))$

Pole Location(s) correspond to the type of signal.

Euler's equations $\cos(\omega\cdot t) = \dfrac{e^{j\cdot\omega\cdot t} + e^{-j\cdot\omega\cdot t}}{2}$ $\qquad \sin(\omega\cdot t) = \dfrac{e^{j\cdot\omega\cdot t} - e^{-j\cdot\omega\cdot t}}{2\cdot j}$

Ex. 4 $f(t) = u(t)\cdot\cos(\omega\cdot t)$

$$F(s) = \int_0^\infty \cos(\omega\cdot t)\cdot e^{-s\cdot t}\,dt = \int_0^\infty \left(\frac{e^{j\cdot\omega\cdot t} + e^{-j\cdot\omega\cdot t}}{2}\right)\cdot e^{-s\cdot t}\,dt = \int_0^\infty \frac{e^{(j\cdot\omega - s)\cdot t} + e^{-(j\cdot\omega + s)\cdot t}}{2}\,dt$$

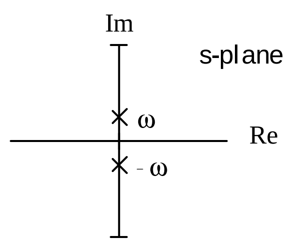
$$= \frac{1}{2}\cdot\int_0^\infty e^{(j\cdot\omega - s)\cdot t}\,dt + \frac{1}{2}\cdot\int_0^\infty e^{-(j\cdot\omega + s)\cdot t}\,dt$$

$$= \frac{1}{2}\cdot\left(\frac{1}{j\cdot\omega - s}\right)\cdot e^{(j\cdot\omega - s)\cdot t}\Bigg|_0^\infty + \frac{1}{2}\cdot\left[\frac{1}{-(j\cdot\omega + s)}\right]\cdot e^{-(j\cdot\omega + s)\cdot t}\Bigg|_0^\infty$$

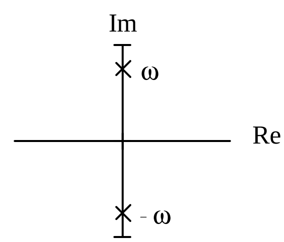
$$= 0 - \frac{1}{2}\cdot\left(\frac{1}{j\cdot\omega - s}\right)\cdot(1) + 0 - \frac{1}{2}\cdot\left[\frac{1}{-(j\cdot\omega + s)}\right]\cdot(1) = \frac{-1}{-2\cdot j\cdot\omega - 2\cdot s} + \frac{-1}{2\cdot j\cdot\omega - 2\cdot s}$$

$$= \frac{1}{2\cdot j\cdot\omega + 2\cdot s} + \frac{-1}{2\cdot\omega - 2\cdot s} = \frac{(2\cdot j\cdot\omega - 2\cdot s) - (2\cdot j\cdot\omega + 2\cdot s)}{(2\cdot j\cdot\omega + 2\cdot s)\cdot(2\cdot j\cdot\omega - 2\cdot s)}$$

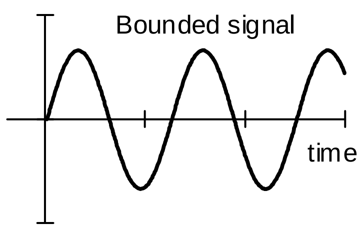
$$= \frac{-4\cdot s}{(2\cdot j\cdot\omega + 2\cdot s)\cdot(2\cdot j\cdot\omega - 2\cdot s)} = \frac{-4\cdot s}{4\cdot j^2\cdot\omega^2 - 4\cdot s^2} = \frac{-s}{-\omega^2 - s^2} = \frac{s}{\omega^2 + s^2}$$



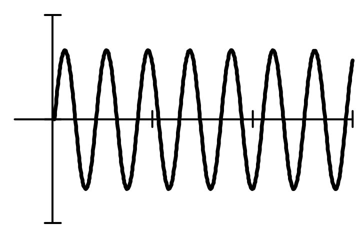
$$\frac{A\cdot s + B}{\left(s^2 + \omega^2\right)}$$



time-domain

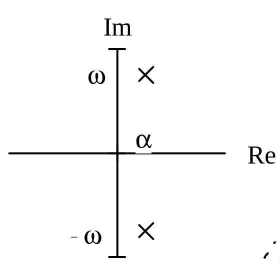


Doesn't converge

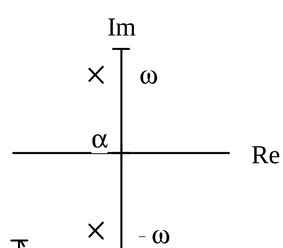


time

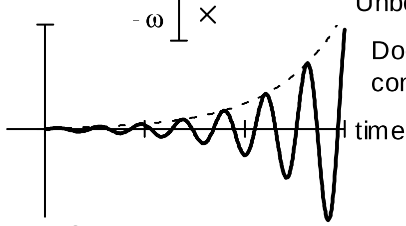
What if the poles have a real component? $\quad f(t) = u(t)\cdot e^{\sigma\cdot t}\cdot\sin(\omega\cdot t)$



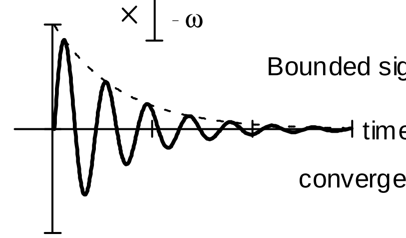
$$\frac{A\cdot s + B}{(s - \alpha)^2 + \omega^2}$$



Unbounded signal



Doesn't converge



Bounded signal

converges

Ex. 5 Multiply by time property

$$f(t) = u(t)\cdot t\cdot e^{a\cdot t} \qquad F(s) = \int_0^{\infty} t\cdot e^{a\cdot t}\cdot e^{-s\cdot t}\,dt = \int_0^{\infty} t\cdot e^{(a-s)\cdot t}\,dt$$

Remember integration by parts:

$$\int h(t)\cdot\frac{d}{dt}g(t)\,dt = h(t)\cdot g(t) - \int g(t)\cdot\frac{d}{dt}h(t)\,dt$$

choose: $h(t) = t$ from which: $\frac{d}{dt}h(t) = 1$

and: $\frac{d}{dt}g(t) = e^{(a-s)\cdot t}$ from which: $g(t) = \int e^{(a-s)\cdot t}\,dt = \frac{e^{(a-s)\cdot t}}{(a-s)}$

$$h(t)\cdot g(t) \quad - \int g(t)\cdot\frac{d}{dt}h(t)\,dt$$

$$F(s) = \int_0^{\infty} t\cdot e^{(a-s)\cdot t}\,dt = t\cdot\frac{e^{(a-s)\cdot t}}{(a-s)}\Bigg|_0^{\infty} - \int_0^{\infty}\frac{e^{(a-s)\cdot t}}{(a-s)}\cdot(1)\,dt = t\cdot\frac{e^{(a-s)\cdot t}}{(a-s)}\Bigg|_0^{\infty} - \frac{e^{(a-s)\cdot t}}{(a-s)^2}\Bigg|_0^{\infty}$$

$$= 0 - 0 - \left[0 - \frac{1}{(a-s)^2}\right]$$

$$= \frac{1}{(a-s)^2} = \frac{1}{(s-a)^2}$$

The easy way:

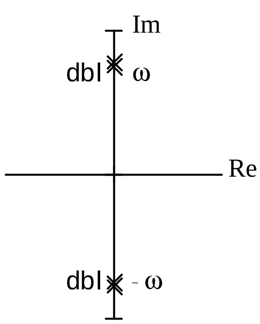
Use the "multiplication by time" property # 5 on p.8 of the Bodson textbook

$$t\cdot x(t) \quad \Longleftrightarrow \quad -\frac{d}{ds}X(s)$$

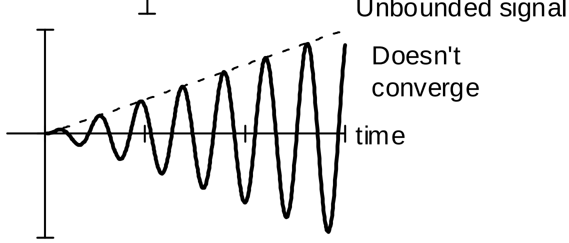
$$t\cdot e^{a\cdot t} \quad \Longleftrightarrow \quad -\frac{d}{ds}\left(\frac{1}{s-a}\right) = -\frac{d}{ds}\left[(s-a)^{-1}\right] = -\frac{1}{-1}\cdot\frac{1}{(s-a)^2}\cdot\left[\frac{d}{ds}(s-a)\right] = \frac{1}{(s-a)^2}\cdot 1 = \frac{1}{(s-a)^2}$$

Anything that works for exponentials also works for sines and cosines...

$$\frac{A\cdot s + B}{\left(s^2+\omega^2\right)^2}$$

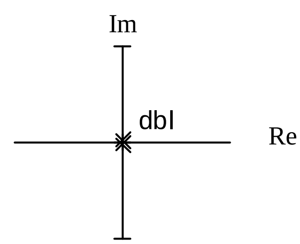


$t\cdot\sin(\omega\cdot t)\cdot u(t)$

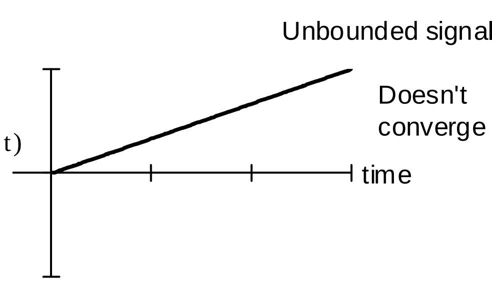


And "DC" too...

$$\frac{A}{s^2}$$

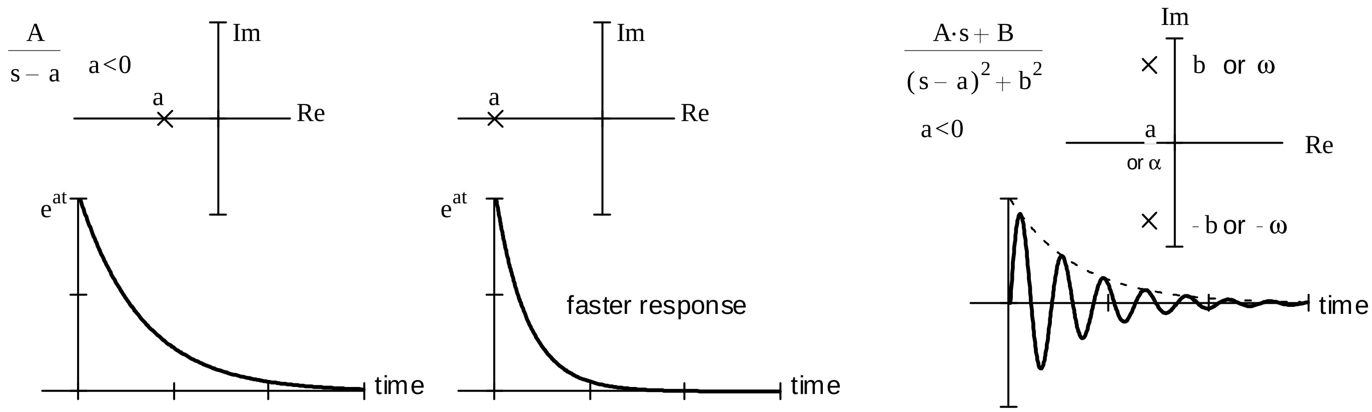


$t\cdot u(t)$



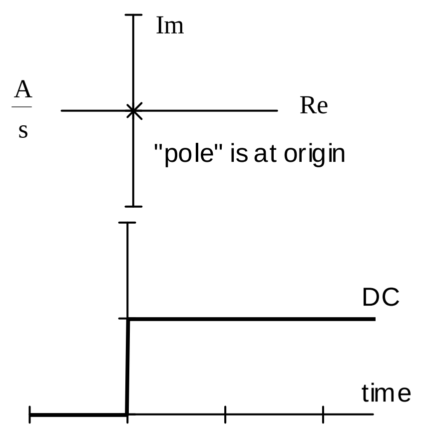
Signal Type, Boundedness, and Convergence can be predicted from the poles

Poles in the Open-Left-Half-Plane (OLHP) Real part of pole is negative $\mathrm{Re}(s_p) < 0$

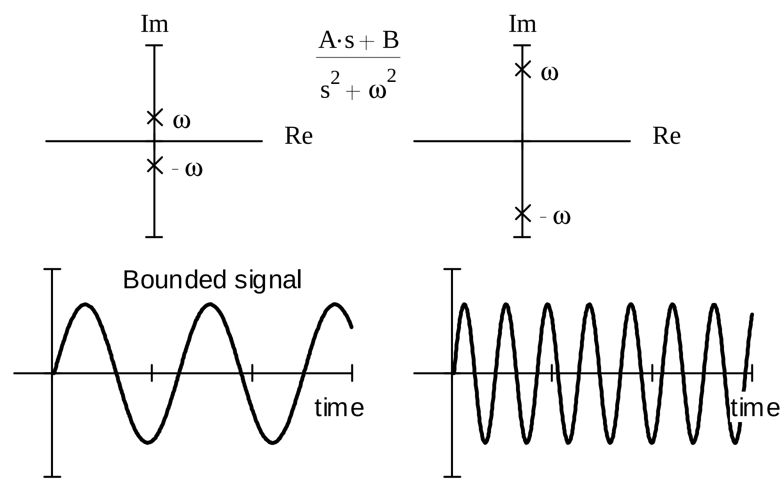


Bounded signals, Converge to zero

Single Poles on Imaginary Axis Real part of pole is zero $\mathrm{Re}(s_p) = 0$

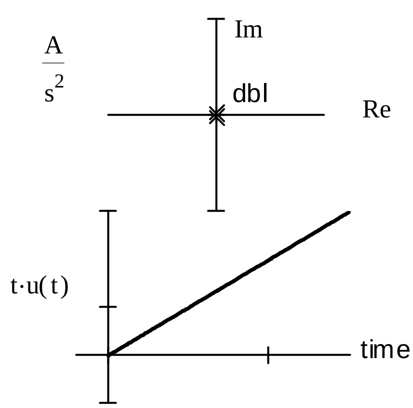


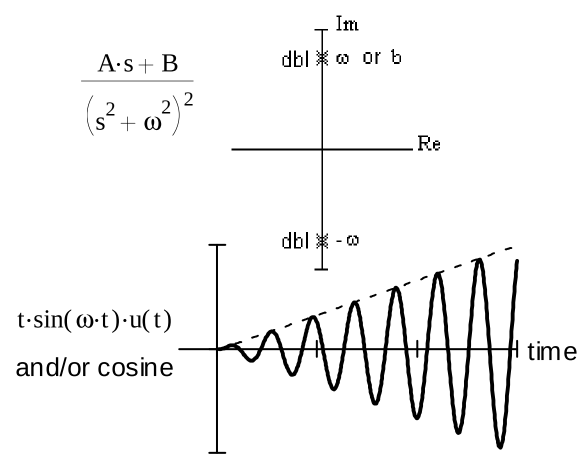
Bounded signal,
Converges to DC value



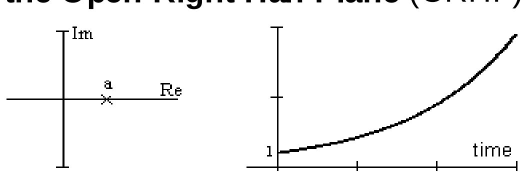
Bounded signals, Don't Converge

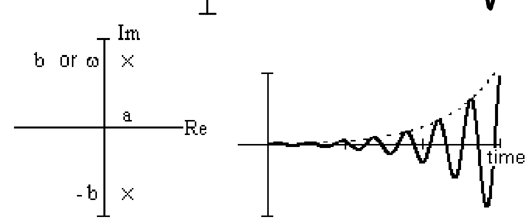
Double Poles on Imaginary Axis or





in the Open-Right-Half-Plane (ORHP)





Unbounded signals, Don't Converge