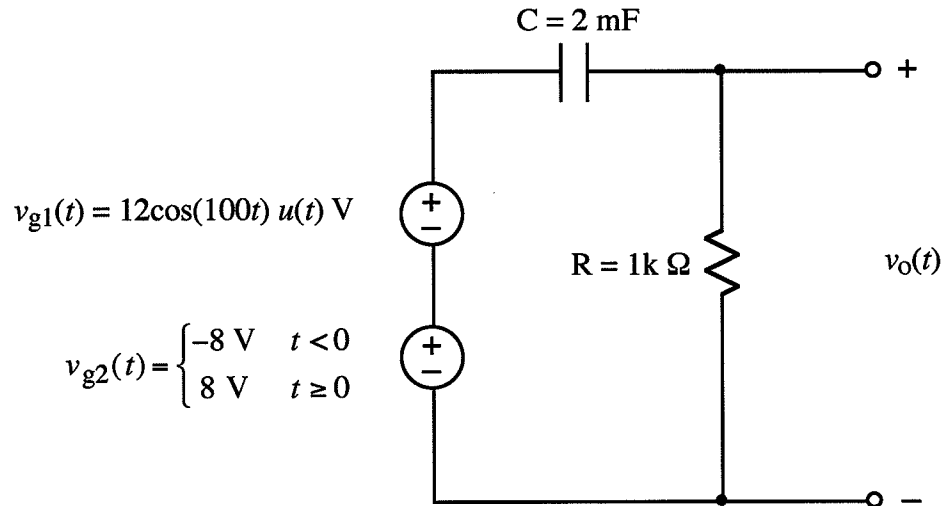


Ex:



- Write the Laplace transform,  $V_{g2}(s)$ , of  $v_{g2}(t)$ .
- Draw the s-domain equivalent circuit, including sources  $V_{g1}(s)$  and  $V_{g2}(s)$ , components, initial conditions for  $C$ , and terminals for  $V_o(s)$ .
- Write an expression for  $V_o(s)$ .
- Apply the final value theorem to find  $\lim_{t \rightarrow \infty} v_o(t)$ .

sol'n: a) We only consider  $v_{g2}(t)$  for  $t > 0$ .

$$V_{g2}(s) = \mathcal{L} \{ 8 \} \text{ V} = \mathcal{L} \{ 8u(t) \} \text{ V} = \frac{8}{s} \text{ V}$$

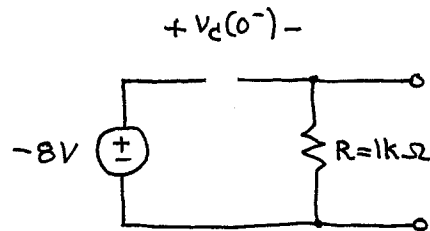
$$\begin{aligned} \text{b) } \mathcal{L} \{ v_{g1}(t) \} &= \mathcal{L} \{ 12\cos(100t)u(t) \text{ V} \} \\ &= 12 \frac{s}{s^2 + 100^2} \text{ V} \end{aligned}$$

Initial conditions on  $C$ :

At  $t = 0^-$ ,  $C$  acts like open circuit.

$$v_{g1}(t < 0) = 0V$$

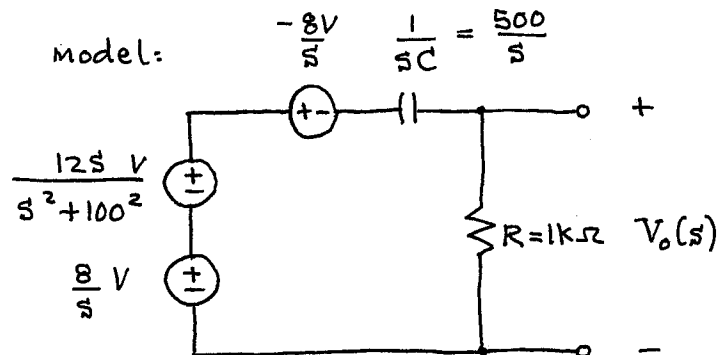
$$v_{g2}(t < 0) = -8V$$



Since no current flows, there is no V-drop across R.

$$\therefore v_c(0^-) = -8V$$

Using a series V source for initial conditions on the C is convenient.



c) We get  $V_o(s)$  from a V-divider formula:

$$V_o(s) = \left( \frac{8V}{s} + \frac{125V}{s^2 + 100^2} - \frac{8V}{s} \right) \frac{R}{R + \frac{1}{sC}}$$

$$V_o(s) = \left( \frac{16V}{s} + \frac{125}{s^2 + 100^2} \right) \frac{s}{s + \frac{1}{RC}}$$

d) We may apply the final value theorem only if the poles of  $V_o(s)$ , with exception of a pole at the origin (of first order), lie in the left half-plane.

Thus, we may not apply the theorem, since  $\frac{12s}{s^2+100^2}$  gives us poles at  $s = \pm j100$  r/s.

This makes sense since  $v_{g1}(t)$  is a sinusoid that never decays. We could solve for  $v_o(t \rightarrow \infty)$  by considering a phasor solution. We would find that  $v_o(t)$  is a sinusoid of frequency 100 r/s that never decays. Thus, there is no unique value for  $v_o(t \rightarrow \infty)$ .