

(Due by Nov. 16 by 6pm in homework locker)

1. (a) The triangular signal of Fig. 1(a) is input to the class A amplifier of Fig. 1(b). The amplifier has the following parameters: $V_{CC}=5V$, $V_{BE}=0.7V$ (assume constant), $V_{CE,sat}=0.3V$, $R_L=2k\Omega$, and the bias current is designed to be $I=V_{CC}/R_L$. Sketch the output voltage v_o .

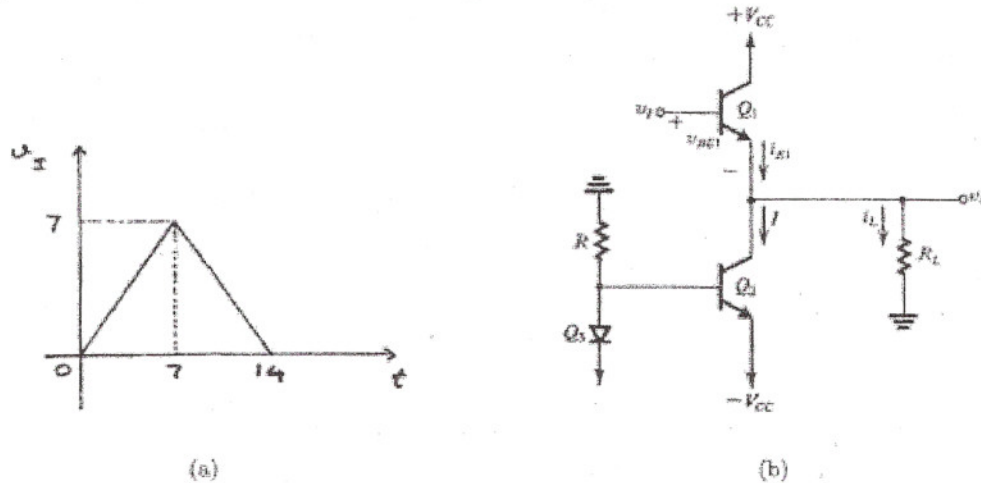
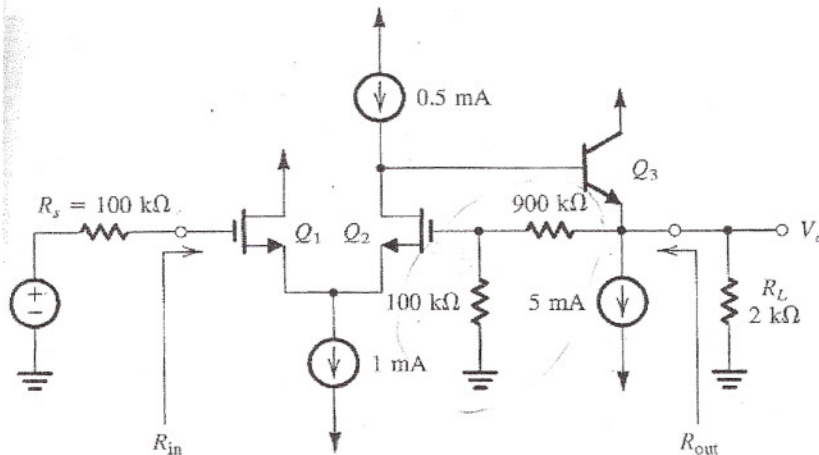


Fig. 1 (a) Input, (b) Output Stage

(b) What type of feedback is employed in the circuit below:



(c) We discussed three types of power amplifiers or output stages in class: **Class A**, **Class B**, and **Class AB**.

Which type has the worst distortion? _____

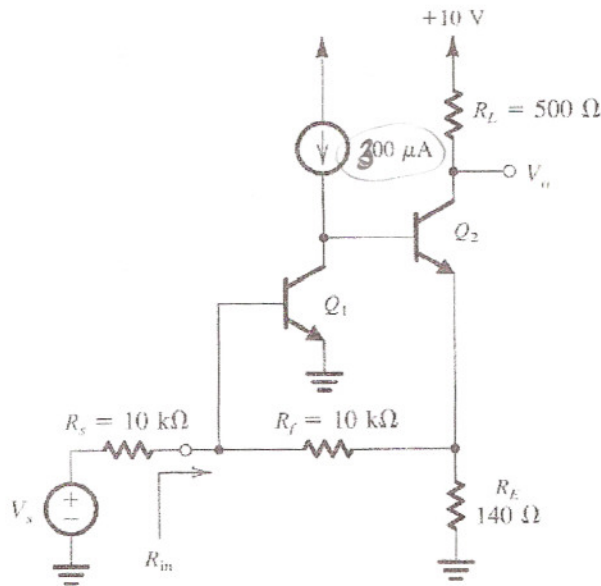
Which type has the highest power efficiency? _____

Which type has the lowest power efficiency? _____

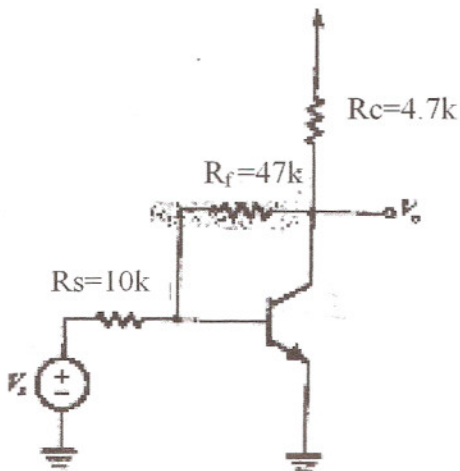
(d) If we put a square wave into a feedback amplifier, and the output shows too much ringing for our application, is the phase margin too high or too low? _____

2. An amplifier has a dc gain of 10^5 and high-frequency poles at 10^5 Hz, 3.16×10^5 Hz, and 10^6 Hz. For a phase margin of 45° find the value of β and the corresponding closed-loop gain.

3. Consider the amplifier below. When the source voltage V_s has zero dc component, the output dc voltage is $V_o = 5V$. Let both BJTs have $\beta = 50$, and assume constant $V_{BE} = 0.7V$ for both BJTs. Determine the dc voltages at all nodes and the dc emitter currents of Q_1 and Q_2 . Use feedback analysis to find V_o/V_s and R_{in} .



4. Consider the circuit shown below. The transistor has $\beta = 100$, and is biased so that the dc operating point is $I_B = 0.015mA$, $I_C = 1.5mA$, and $V_o = 4.7V$. Determine the small-signal loop gain.



5. Consider a feedback amplifier for which the open-loop gain $A(s)$ is given by

$$A(s) = \frac{1000}{(1 + s/10^4)(1 + s/10^5)^2}$$

If the feedback factor β is independent of frequency, find the frequency at which the phase shift is 180° , and find the critical value of β at which oscillations will commence. (Hint: A good initial guess for the desired frequency is 10^5 rad/sec.)

6. The datasheet of the PN2222A bipolar transistor contains the following information:

Maximum allowable junction temperature $T_{Jmax} = 150^{\circ}\text{C}$

Maximum power dissipation $P_{Dmax} = 625 \text{ mW}$ at ambient temperature $T_A = 25^{\circ}\text{C}$

Junction-to-case thermal resistance $\theta_{JC} = 80^{\circ}\text{C/W}$

(a) Draw an equivalent electrical circuit for this thermal system. Label all components as well as the junction temperature T_J , the case temperature T_C , and the ambient temperature T_A . (Note that we are not using a heat sink in this problem.)

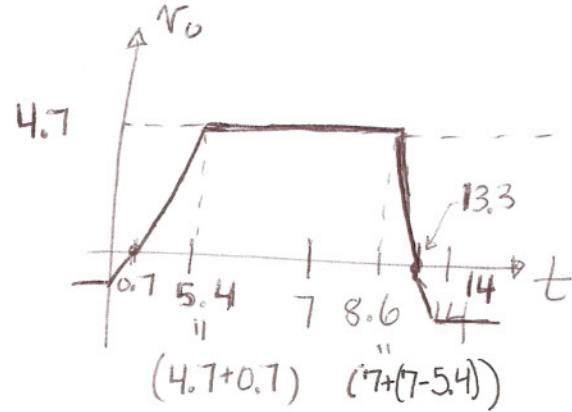
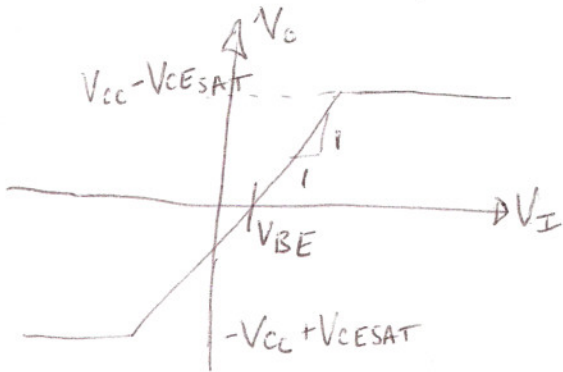
(b) Find the case-to-ambient thermal resistance θ_{CA} .

(c) Assuming an ambient temperature of 25°C , what is the case temperature T_C when the transistor dissipates 500 mW?

7. Exercises 12.31 and 13.18

HW #10 sol'n

1. a. Class A amplifier transfer characteristic



b. Series-shunt

c. worst distortion: Class B

highest power efficiency: Class B (or AB is almost the same)

lowest power efficiency: Class A.

d. too low

$$2. A(s) = \frac{10^5}{\left(\frac{s}{10^5} + 1\right) \left(\frac{s}{3.16 \times 10^5} + 1\right) \left(\frac{s}{10^6} + 1\right)}$$

$$\text{pm} = 45^\circ \Rightarrow \tan^{-1}\left(\frac{f}{10^5}\right) + \tan^{-1}\left(\frac{f}{3.16 \times 10^5}\right) + \tan^{-1}\left(\frac{f}{10^6}\right) = 135^\circ$$

$$\therefore f \cong 3.16 \times 10^5 \text{ Hz}$$

$$|A \cdot \beta| = 1 = \frac{10^5 \cdot \beta}{\sqrt{\left(\frac{3.16 \times 10^5}{10^5}\right)^2 + 1} \sqrt{2} \sqrt{\left(\frac{3.16 \times 10^5}{10^6}\right)^2 + 1}}$$

$$\Rightarrow \beta \cong \boxed{49 \mu}$$

$$A_f = \frac{10^5}{1 + (10^5)(49 \mu)} \cong \boxed{17 \text{ K}}$$

3. (cont.)

$$g_{m1} = \frac{I_{c1}}{V_T} = \frac{100\mu}{25m} = 4m \text{ A/V} \Rightarrow r_{\pi1} = \frac{\beta}{g_{m1}} = \frac{50}{4m} = 12.5k\Omega$$

$$g_{m2} = \frac{I_{c2}}{V_T} = \frac{10m}{25m} = 400m \text{ A/V} \Rightarrow r_{\pi2} = \frac{50}{400m} = 20\Omega$$

$$I_o = g_{m2} V_2$$

$$V_2 = -\underbrace{g_{m1} V_1 (r_{\pi2} + 138(\beta+1))}_{V_{b2}} - \underbrace{g_{m2} V_2 (138 \parallel \frac{r_{\pi2}}{\beta+1})}_{V_{e2}}$$

$$V_2 = -28.152 V_1 - 0.16 V_2 \Rightarrow V_2 \triangleq -24.4 V_1$$

$$V_1 = I_i (10k \parallel 10.14k \parallel r_{\pi1}) \triangleq (3589) I_i$$

$$A \triangleq \frac{I_o}{I_i} = 400m (-24.4) (3589) \triangleq -35k \text{ A/A}$$

$$A_T = \frac{A}{1+A\beta} = \frac{-35k}{1+(-35k)(-0.0138)} \triangleq \boxed{-73 \text{ A/A}}$$

$$V_o = -I_o (500) \quad \text{and} \quad I_s = \frac{V_s}{10k}$$

$$\frac{V_o}{V_s} : \frac{I_o}{I_s} = \frac{-V_o}{500} \cdot \frac{10k}{V_s} \Rightarrow \frac{V_o}{V_s} = 73 \cdot \frac{500}{10k} \triangleq \boxed{3.65 \text{ V/V}}$$


$$R_i = (10k \parallel 10.14k \parallel r_{\pi1}) \approx 3589$$

$$R_{if} = \frac{R_i}{1+A\beta} \approx \boxed{64\Omega}$$

4. shunt-shunt configuration:

• feedback \Rightarrow 

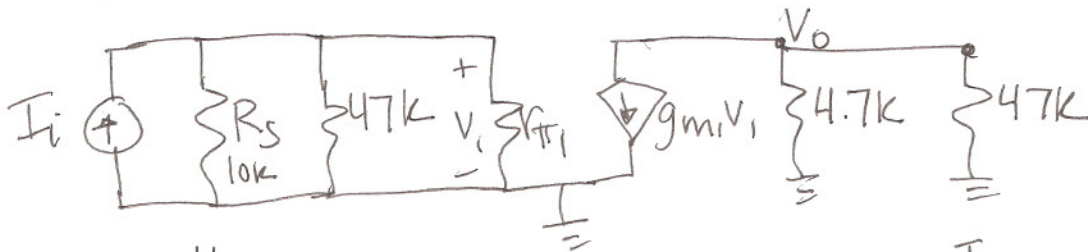
$$R_{11} = R_{22} = R_f = 47k$$

$$\beta \equiv \frac{I_f}{V_o}$$


$$V_o = -I_f \cdot R_f$$

$$\therefore \beta = -\frac{1}{R_f} = -\frac{1}{47k}$$

A-circuit



$$A \equiv \frac{V_o}{I_i}$$

$$V_o = -g_{m1} V_i (4.7k \parallel 47k)$$

$$V_i = I_i (10k \parallel 47k \parallel r_{\pi 1})$$

$$g_{m1} = \frac{I_{c1}}{V_T} = \frac{1.5m}{25m} = 60m$$

$$r_{\pi 1} = \frac{\beta}{g_{m1}} = \frac{100}{60m} \approx 1.7k$$

$$A \equiv \frac{V_o}{I_i} \approx -60m (4.3k)(1.410) = -363,780$$

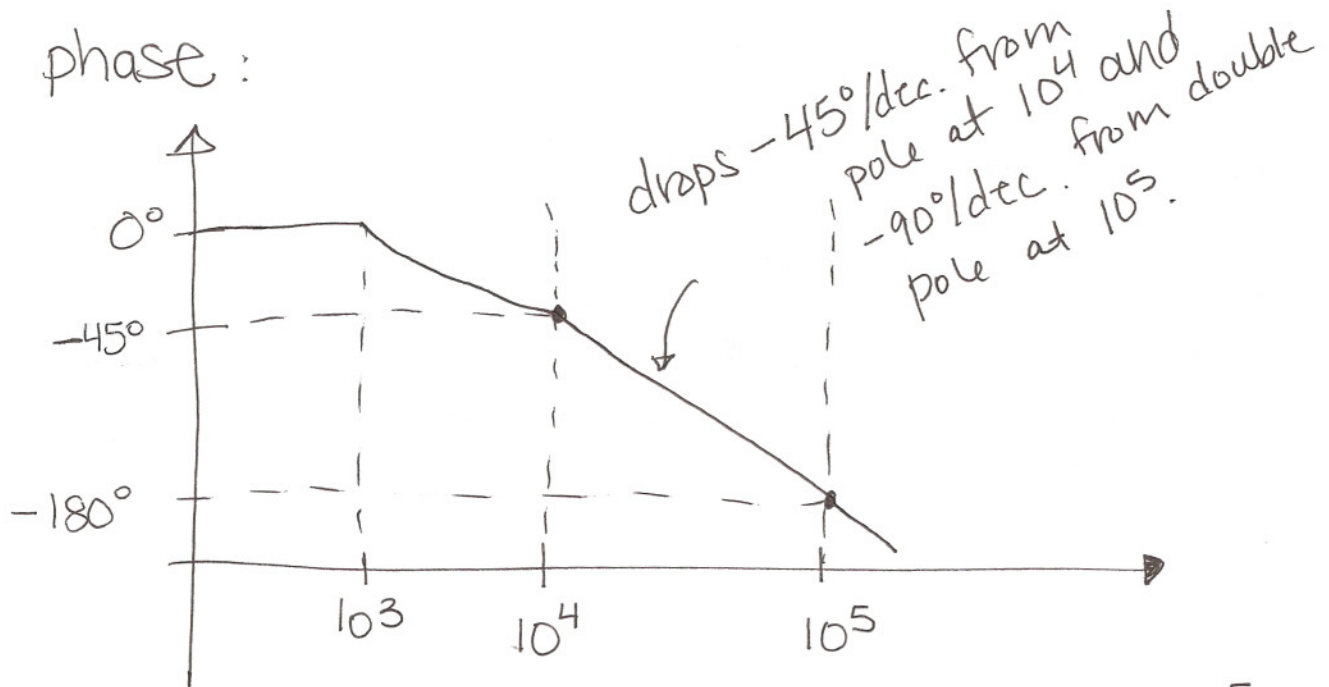
$$\frac{V_o}{I_s} = A_f = \frac{-363,780}{1 + (-363,780)(-\frac{1}{47k})} \approx \boxed{-42k \text{ V/A}}$$

converting to $V_s \Rightarrow$

$$I_s = \frac{V_s}{10k} \Rightarrow -42k = \frac{V_o}{I_s} = \frac{V_o \cdot 10k}{V_s}$$

$$\frac{V_o}{V_s} = \boxed{-4.2 \text{ V/V}}$$

5. phase :



\therefore phase shift at 180° should be around $10^5 \Rightarrow$

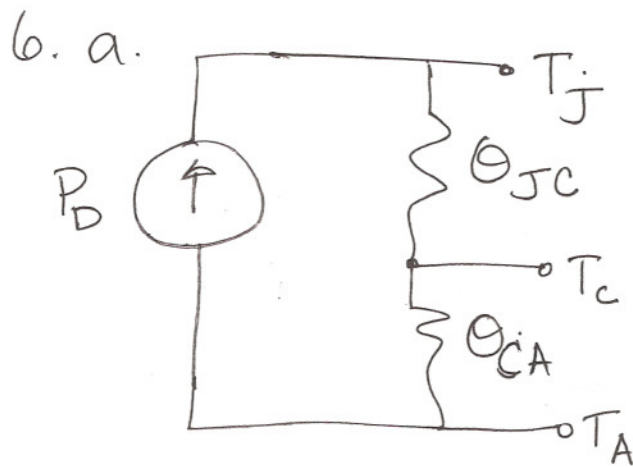
$$\angle A(j\omega) = \tan^{-1}\left(\frac{\omega}{10^4}\right) - 2 \tan^{-1}\left(\frac{\omega}{10^5}\right) = -180^\circ$$

$$\omega \approx \boxed{1.1 \times 10^5 \text{ rad/sec.}}$$

Oscillations commence when $|AB|=1$ and the phase from A is at 180°

$$\therefore \frac{1,000 \cdot B}{\sqrt{\left(1 + \left(\frac{1.1 \times 10^5}{10^4}\right)^2\right) \left[1 + \left(\frac{1.1 \times 10^5}{10^5}\right)^2\right]}} \geq 1$$

$$B \geq \boxed{0.0244}$$



b. at $T_A = 25^\circ\text{C}$ then $P_D = 625\text{mW}$

$$\theta_{JC} = 80^\circ\text{C/W}$$

$$T_J = 150^\circ\text{C}$$

$$\therefore P_D (\theta_{JC} + \theta_{CA}) = T_J - T_A$$

$$\theta_{CA} = \frac{T_J - T_A}{P_D} - \theta_{JC} = \frac{(150 - 25)}{625\text{m}} - 80^\circ\text{C/W}$$

$$\theta_{CA} = \boxed{120^\circ\text{C/W}}$$

$$\text{OR } T_J - T_c = P_D (\theta_{JC}) \Rightarrow T_c = 150^\circ - 625\text{m}(80^\circ) = 100^\circ$$

$$T_c - T_A = P_D (\theta_{CA}) \Rightarrow \frac{100 - 25}{625\text{m}} = \theta_{CA} = 120^\circ\text{C/W}$$

c. $P_D = 500\text{mW} \Rightarrow$

$$T_c = T_A + P_D (\theta_{CA}) = 25 + 500\text{m}(120^\circ) = 85^\circ$$

12.31

From Eq. (12.96) & (12.97)

$$\begin{aligned}C_3 = C_4 &= \omega_0 T_c C \\ &= 2\pi 10^4 \times \frac{1}{200 \times 10^3} \times 20 \\ &= \underline{\underline{6.283 \text{ pF}}}\end{aligned}$$

From Eq. (12.99)

$$C_5 = \frac{C_4}{Q} = \frac{6.283}{20} = \underline{\underline{0.314 \text{ pF}}}$$

From Eq. (12.100)

$$\text{Centre-frequency gain} = \frac{C_6}{C_5} = 1$$

$$C_6 = C_5 = \underline{\underline{0.314 \text{ pF}}}$$

13.18

To obtain a triangular waveform with 10-V peak-to-peak amplitude we should have

$$V_{TH} = -V_{TL} = 5V$$

$$\text{But } V_{TL} = -L + \frac{R_1}{R_2}$$

$$\text{Thus } -5 = -10 + \frac{10}{R_2}$$

$$R_2 = \underline{\underline{20 \text{ k}\Omega}}$$

For 1 kHz frequency, $T = 1 \text{ ms}$.

$$\text{Thus, } \frac{T}{2} = 0.5 \times 10^{-3} = CR \frac{V_{TH} - V_{TL}}{L}$$

$$= 0.01 \times 10^{-6} \times R \times \frac{10}{10}$$

$$R = \underline{\underline{50 \text{ k}\Omega}}$$