Ex:



Given the resistor connected as shown and using not more than one each $R, L$, and $C$ in the dashed-line box, design a circuit to go in the dashed-line box that will produce the band-pass $|\mathrm{H}(j \omega)|$ vs. $\omega$ shown above. That is:

$$
\max _{\omega}|H(j \omega)|=\frac{1}{4} \text { and occurs at } \omega_{0}=10 \mathrm{Mr} / \mathrm{s}
$$

The bandwidth, $\beta$, of the filter is $500 \mathrm{kr} / \mathrm{s}$.

$$
|H(j \omega)|=0 \text { at } \omega=0 \text { and } \quad \lim _{\omega \rightarrow \infty}|H(j \omega)|=0
$$

SOL'N: To achieve the peak at $\omega_{0}$, we may use a series LC in the top rail or a parallel LC from the top to bottom rail. To achieve a gain of $1 / 4$ at $\omega_{0}$, we must use a vertical resistance to form a $V$-divider with $R$.


We have two possible circuit configurations:


Series configuration


Parallel Configuration

Note: $R_{1}$ must be to the right of $L$ and $C$ in order to have any effect from the $L$ and $C$ in the series configuration.

For the series configuration, the $L$ and $C$ are to act like a wire at $\omega_{0}=10 \mathrm{M} \mathrm{r} / \mathrm{s}$.

$$
j \omega_{0} L+\frac{1}{j \omega_{0}^{C}}=0 \quad \text { or } \quad \omega_{0}^{2}=(10 M r / s)^{2}=\frac{1}{L C}
$$

The bandwidth when using a series $L$ and $C$ is

$$
\beta=\frac{R_{0}}{L}=500 \mathrm{kr} / \mathrm{s}
$$

To determine the value of $R_{\text {Eg }}$, we view the filter as a, standard RLC filter and a V-divider.


$$
\begin{aligned}
& H(j \omega)=\frac{V_{0}^{\prime}}{V_{i}^{\prime}} \cdot \frac{V_{0}}{V_{0}^{\prime}} \\
& H(j \omega)=H^{\prime}(j \omega) \cdot \frac{R_{1}}{R+R_{1}}
\end{aligned}
$$

The cutoff frequencies for $H(j \omega)$ are the same as the cutoff frequencies for $H^{\prime}(j \omega)$ :

$$
\omega_{C 1,2}= \pm \frac{R_{e 8}+\sqrt{2 L} \sqrt{\left(\frac{R_{e 8}}{2 L}\right)^{2}+\omega_{0}^{2}}, \quad \beta=\frac{R_{e g}}{L}, ~}{2}
$$

where $R_{\text {eg }}=R+R_{1}=400 \Omega$
Using $\beta=R_{8} / L$, we find $L$ :

$$
L=\frac{R_{e g}}{\beta}=\frac{400 \mathrm{R}}{500 \mathrm{kr} / 3}=0.8 \mathrm{mH} \text { or } 800 \mu \mathrm{H}
$$

Using $w_{0}^{2}=\frac{1}{L C}$ and $L=800 \mu H_{2}$ we find $C=$

$$
\begin{aligned}
& C=\frac{1}{\omega_{0}^{2} L}=\frac{1}{10 M \cdot 1094 \cdot 800 \mu}=\frac{1 \mu F}{80 K} \\
& C=12.5 \mathrm{pF}
\end{aligned}
$$

Summary of series RLC: $R_{1}=100 \Omega, L=800 \mu H, C=12.5 \mathrm{~F}$

For the parallel configuration, we move $R$, to the left of $L$ and $C$ and use a Theremin equivalent of $V_{i}, R$, and $R_{1}$.


To find the Theremin equivalent, we find $V_{T h}$ by finding the open-circuititpof the $V_{i}, R$, and $R_{t}$ dircait.

$$
\left.V_{i} \stackrel{T}{i}_{R_{1}}^{R_{0}}\right\}_{0}^{R=300 \Omega}+a+
$$

To find $R_{T h}$, we turn off $V_{i}$ and look in from terminals $a$ and $b$. The resistance seen is

$$
R_{T h}=R / / R_{1}=300 \Omega / / 100 \Omega=75 \Omega
$$

Using the filter with the Thevenin equivalent, we have

$$
H(j \omega)=\frac{V_{0}}{V_{i}}=\frac{R_{1}}{R+R_{1}} \frac{V_{0}}{V_{i}^{\prime}}=\frac{1}{4} H^{\prime}(j \omega)
$$

where $H^{\prime}(j \omega) \equiv \frac{V_{0}}{V_{i}}$, where $V_{i}{ }^{\prime} \equiv \frac{R_{1}}{R+R_{1}} V_{i}$
The cutoff frequencies of $H^{\prime}(j w)$ are the same as the cutoff frequencies of $H(j w)$.

$$
\begin{aligned}
\omega_{C 1,2}= \pm \frac{1}{2 R C}
\end{aligned}+\sqrt{\left(\frac{1}{2 R C}\right)^{2}+\omega_{0}^{2}}, \beta=\frac{1}{R C}
$$

Using $R_{\text {rh }}$ and $\beta$, we find $C$ :

$$
C=\frac{1}{R \beta}=\frac{1}{75 \Omega \cdot 500 \mathrm{kr} / \mathrm{s}}=26.6 \mathrm{nF}
$$

Using $L$ and $\omega_{0}^{2}$, we find $L$ :

$$
L=\frac{1}{\omega_{0}^{2} C}=\frac{1}{10 M l 0 M 26.6 n}=375 n H
$$

